



Introduction of Hybrid Multiscale Simulation Technologies

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March 07, 2013

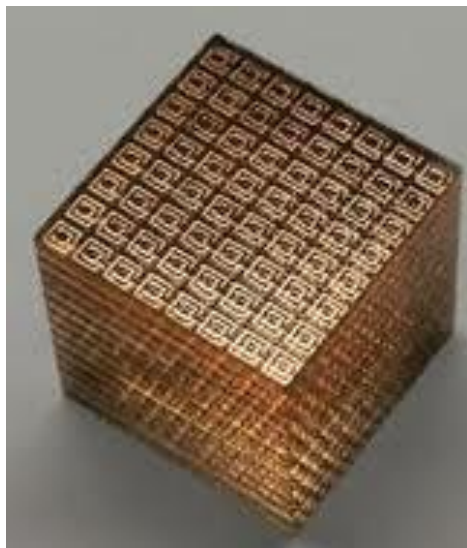
Contents

- ▶ **Motivation and Vision**
- ▶ **Field–Circuit Co–Simulation**
- ▶ **Multiscale Simulation**
- ▶ **About the Company**

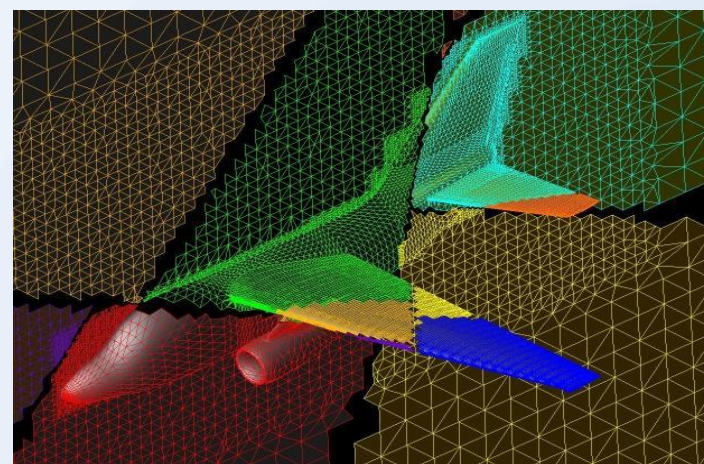
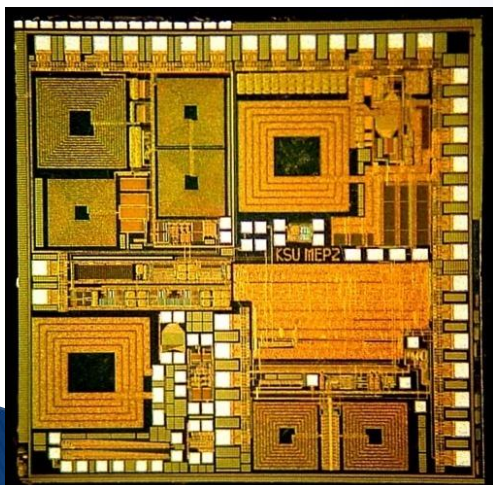
Why Hybrid & Multiscale?

- » Need
- » Challenge
- » Nature
- » Vision

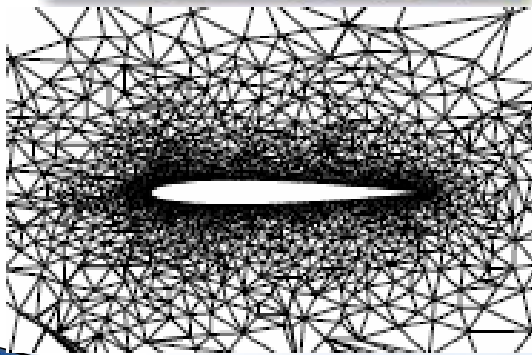
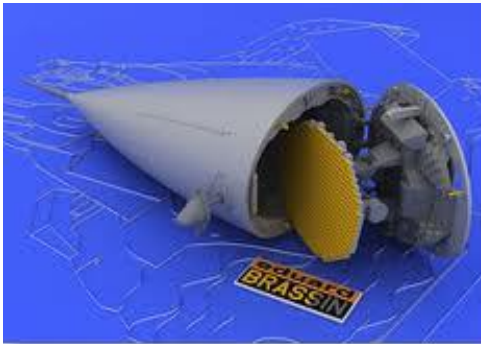
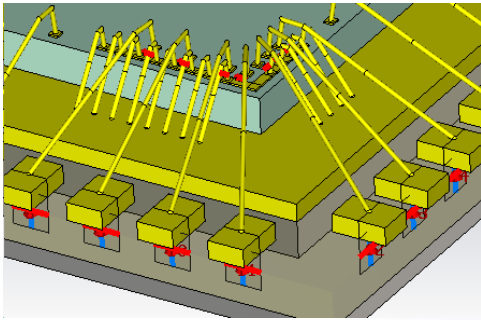
CAD Tools are Targeting More Practical Applications



Engineering
Simulation



Understand the Real Challenge



- ▶ **Small features**
 - Sharp corners
 - Thin wires and surfaces
- ▶ **Multiscale**
 - From meters to millimeters
 - Thousands of small (mm level) units
 - Large (m level) platforms
 - L-band to U-band (1–60GHz)
- ▶ **Multiphysics**
 - Liquid or gas (CFD)
 - Heat transfer (Thermal)
 - RF-Circuit system (MNA)
 - THz antennas (Diffusion)
- ▶ **Solution**
 - Breathe the emerging technologies
 - Hybrid multiscale simulation engine

A Brief Review of Popular Transient Techniques

Integrate powerful solvers



- Hybrid EM-SPICE
- Hybrid FDTD-FETD-SETD
- Hybrid EM-Particle
- Hybrid Explicit-Implicit
- And more ...

- ▶ **FDTD – Finite Difference**
- ▶ **FETD – Finite Element**
- ▶ **SETD – Spectral Element**
- ▶ **DGTD – Discontinuous Galerkin**
- ▶ **SPICE – Simulation Program with Integrated Circuit Emphasis**

NextGen Simulator

The thing is:

For multi-scale, multi-application problems, is hybrid technique simply putting things together?



- ▶ **Single Algorithm, Multi Physics**
 - FEM solvers
 - Mechanics, CFD, CEM
- ▶ **Single Physics, Multi Algorithms**
 - Multiscale EM problems
 - Hybrid time-domain solvers
 - Hybrid freq-domain solvers
- ▶ **Multi Algorithms, Multi Physics**
 - Hybrid FD, FE, FV
 - Hybrid DE, IE
 - Hybrid TD, FD
- ▶ **Multi Process**
 - Hardware computing (FPGA/CPLD)
 - Parallel computing (MPI/GPU/Multithreading)

Field-Circuit Co-Simulation

»» Key Techniques
Applications

Background

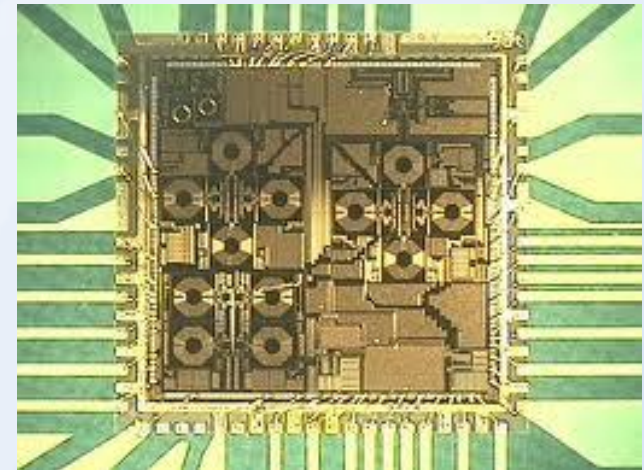
▶ Modern Circuit Systems

- Multifunctional
- High operating frequency
- Large integration scale



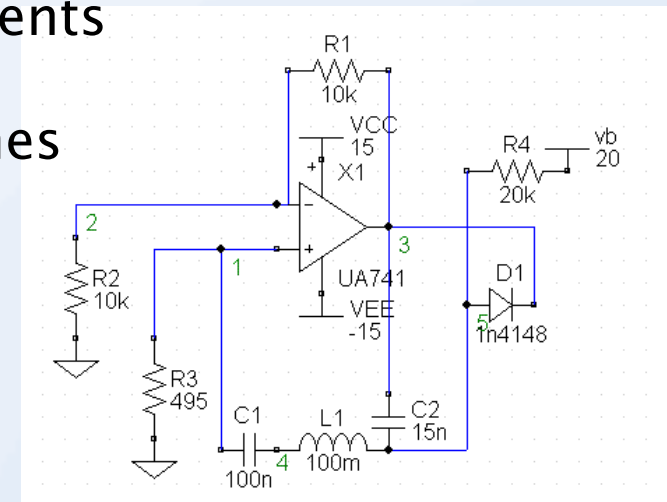
▶ Challenges

- Complex material and device
 - Equivalent circuit modeling is limited
- High frequency
 - EM effects
- Small distance
 - Interference



Circuit Solver – SPICE

- ▶ Modified Nodal Analysis (MNA)
 - General analysis method used to compute nodal voltages and branch currents of a lumped electronic circuit network
 - Graph Theory
 - Circuit is represented via a graph
 - Branches of the graph are circuit elements
 - Every branch is bound by two nodes
 - Nodes form the connectivity of branches
 - ▶ 1 reference node (typically ground)
 - Every node assigned a voltage
 - Every branch assigned a current



MNA

- ▶ DOFs
 - Node voltages
 - Branch currents of:
 - Voltage sources and inductors
- ▶ Linear system of equations
 - Kirchoff's current law (KCL) at every non reference node:
 - KVL about all branches supporting a voltage source

$$\sum_{i=1}^{N_a} I_i = 0 \qquad \sum_{i=1}^{N_b} V_i = 0$$

- ▶ MNA for general circuit
 - Reactive elements (L & C)
 - Non-linear elements

$$\begin{pmatrix} \mathbf{X} & \mathbf{0} \\ \mathbf{0} & \mathbf{0} \end{pmatrix} \begin{pmatrix} \dot{\mathbf{x}}_{sp} \\ \mathbf{i}_v \end{pmatrix} + \begin{pmatrix} \mathbf{Y} & \mathbf{B}^T \\ \mathbf{B} & \mathbf{0} \end{pmatrix} \begin{pmatrix} \mathbf{x}_{sp} \\ \mathbf{i}_v \end{pmatrix} + \begin{pmatrix} \mathbf{i}(\mathbf{x}_{sp}) \\ \mathbf{0} \end{pmatrix} = \begin{pmatrix} -\mathbf{i}_s \\ \mathbf{v}_s \end{pmatrix}$$

Example

► MNA

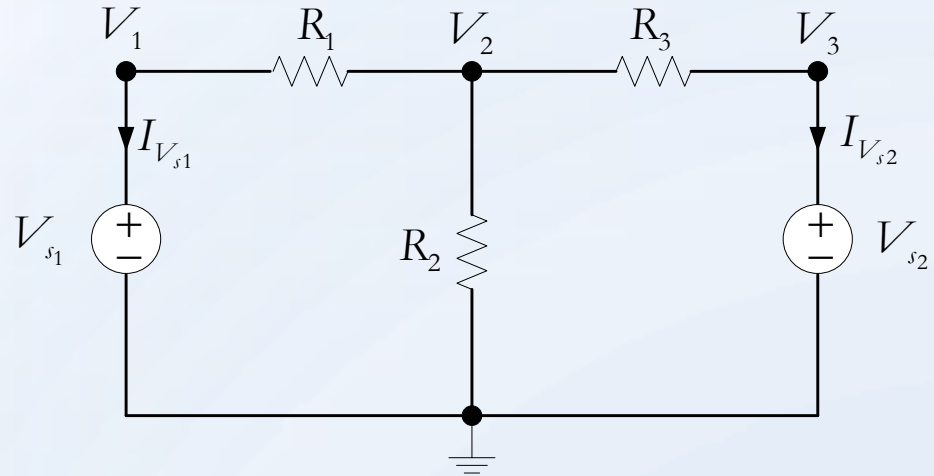
$$\frac{V_1 - V_2}{R_1} + I_{v_{s1}} = 0$$

$$\frac{V_2 - V_1}{R_1} + \frac{V_2}{R_2} + \frac{V_2 - V_3}{R_3} = 0$$

$$\frac{V_3 - V_2}{R_3} + I_{v_{s2}} = 0$$

$$V_1 = V_{s1}$$

$$V_3 = V_{s2}$$



► Linear System

$$\begin{pmatrix} \frac{1}{R_1} & -\frac{1}{R_1} & 0 & 1 & 0 \\ -\frac{1}{R_1} & \frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3} & -\frac{1}{R_3} & 0 & 0 \\ 0 & -\frac{1}{R_3} & \frac{1}{R_3} & 0 & 1 \\ 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \end{pmatrix} \begin{pmatrix} V_1 \\ V_2 \\ V_3 \\ I_{v_{s1}} \\ I_{v_{s2}} \end{pmatrix} = \begin{pmatrix} 0 \\ 0 \\ 0 \\ V_{s1} \\ V_{s2} \end{pmatrix}$$

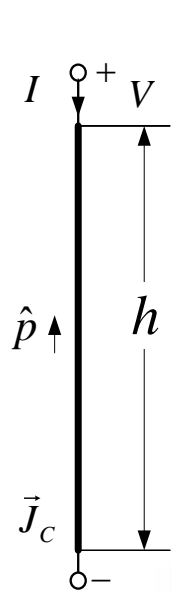
$$\Rightarrow \begin{pmatrix} \mathbf{Y} & \mathbf{B}^T \\ \mathbf{B} & \mathbf{0} \end{pmatrix} \begin{pmatrix} \mathbf{v}_{sp} \\ \mathbf{i}_v \end{pmatrix} = \begin{pmatrix} \mathbf{0} \\ \mathbf{v}_s \end{pmatrix}$$

Field–Circuit Port (Cont'd)

- ▶ **Port Definition**
 - An interface that handles the data exchanging between two solvers
- ▶ **Circuit Port**
 - Convert the local E–field into a circuit voltage and couple the circuit current back into EM solver as a equivalent “J” current source.
- ▶ **Quasi–Static Approximations**
 - E–field is conservative within the port region
 - Line integration of the field is assumed to be independent of the path
 - The feed–back current is evenly distributed through the port region

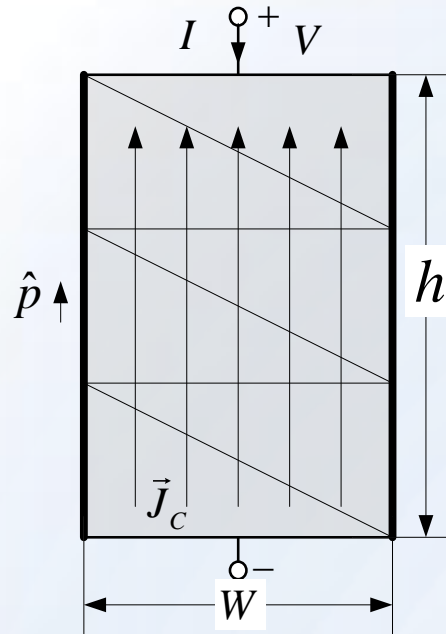
Bo Zhao, Young, J.C, Gedney, S.D "SPICE Lumped Circuit Subcell Model for the Discontinuous Galerkin Finite-Element Time-Domain Method," *Microwave Theory and Techniques, IEEE Transactions on* , vol.60, no.9, pp.2684-2692, Sept. 2012

Field-Circuit Port



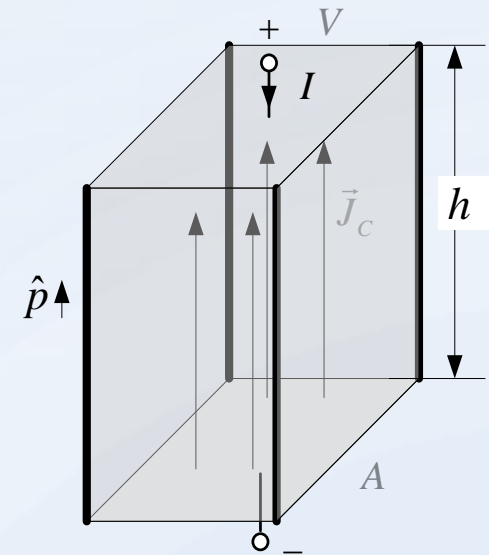
$$V \approx - \int_{L_c} \vec{E} \cdot \hat{p} dl$$

$$\vec{J}_c = \hat{p} I$$



$$V \approx - \frac{1}{W} \int_{S_c} \vec{E} \cdot \hat{p} ds$$

$$\vec{J}_c = \hat{p} \frac{I}{W}$$

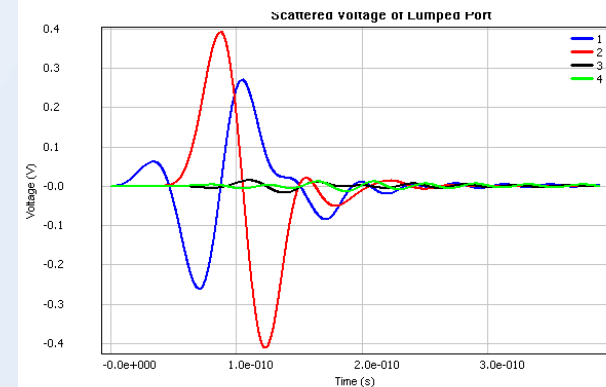
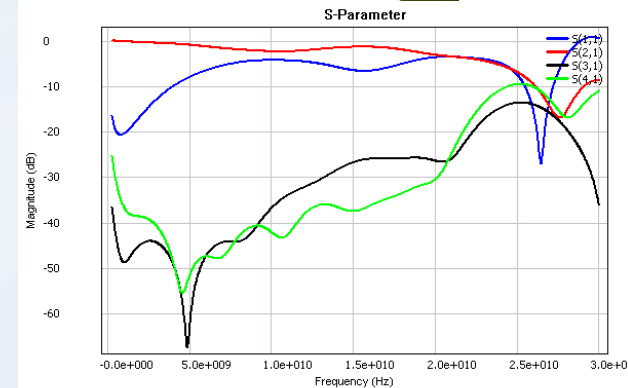
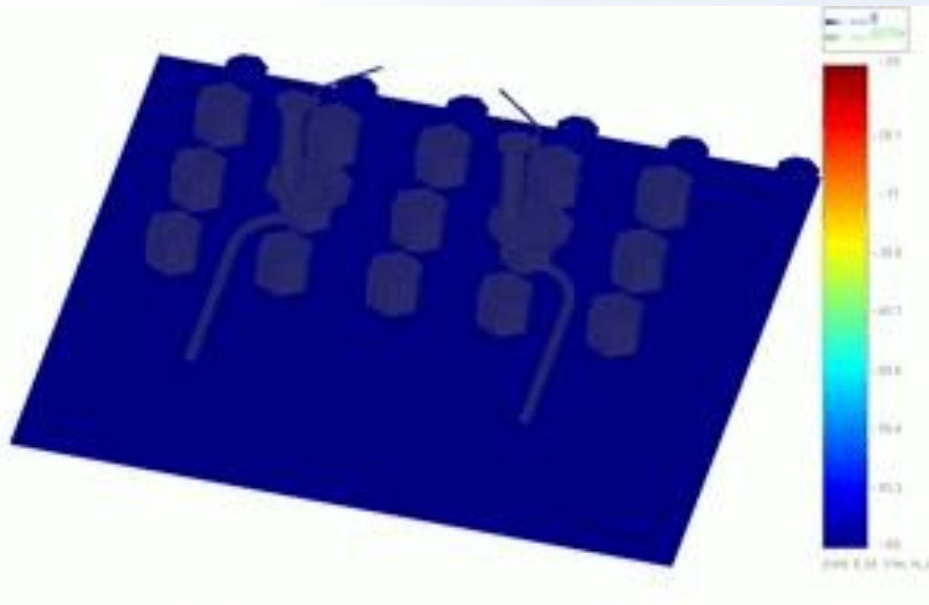
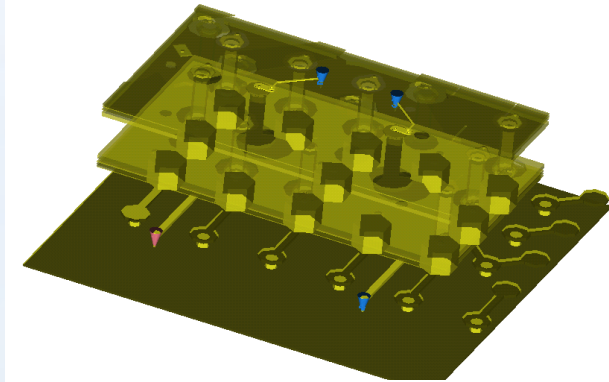


$$V \approx - \frac{1}{A} \int_{V_c} \vec{E} \cdot \hat{p} dv$$

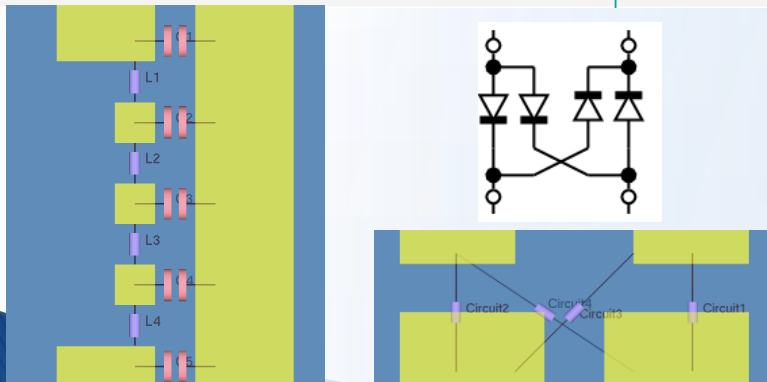
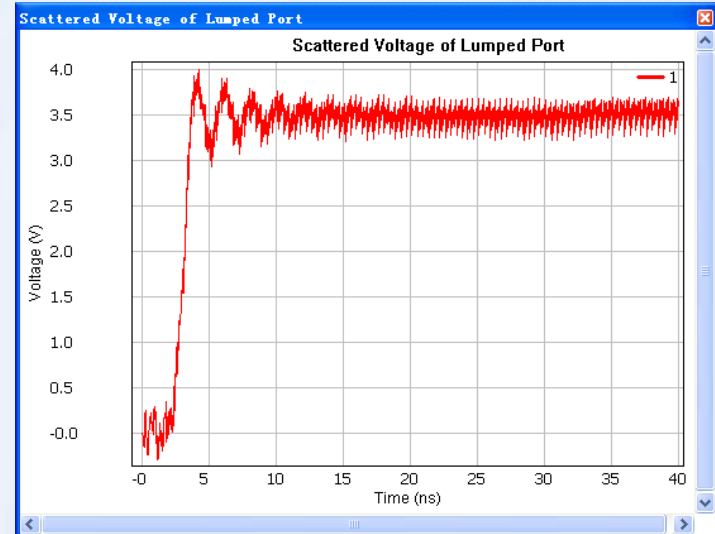
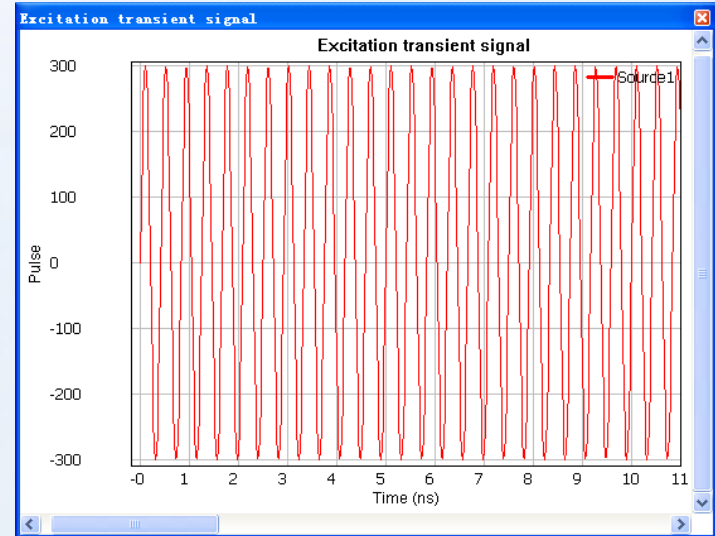
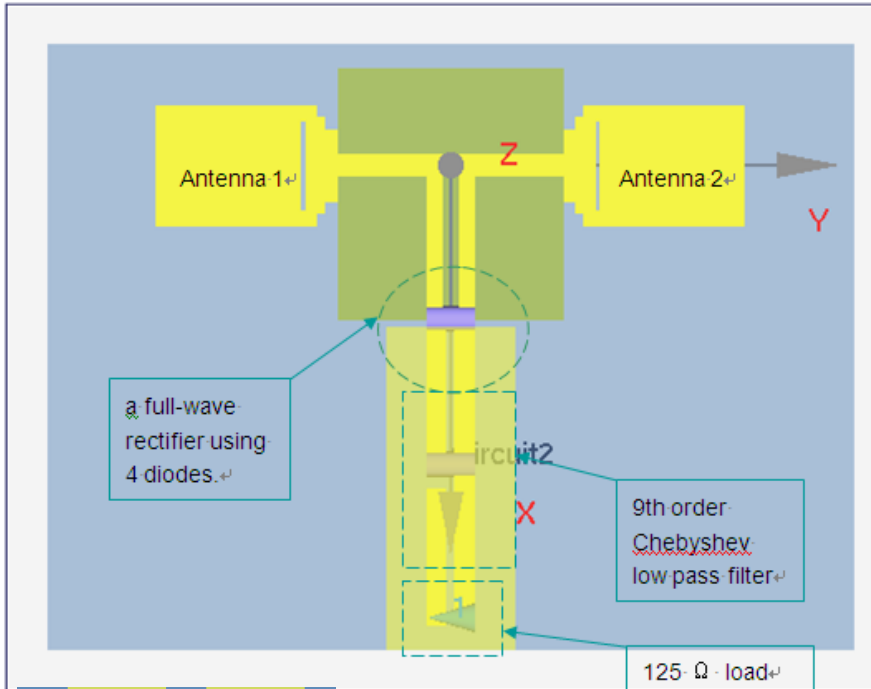
$$\vec{J}_c = \hat{p} \frac{I}{A}$$

Crosstalk Analysis of an IC Package

- ▶ IC package
- ▶ Simulated by Wavenology EM using a quad-core PC. Takes around 10 hours and uses 4GB memory
- ▶ The sink in S11 indicates a resonance, and it is revealed in the surface current snapshot

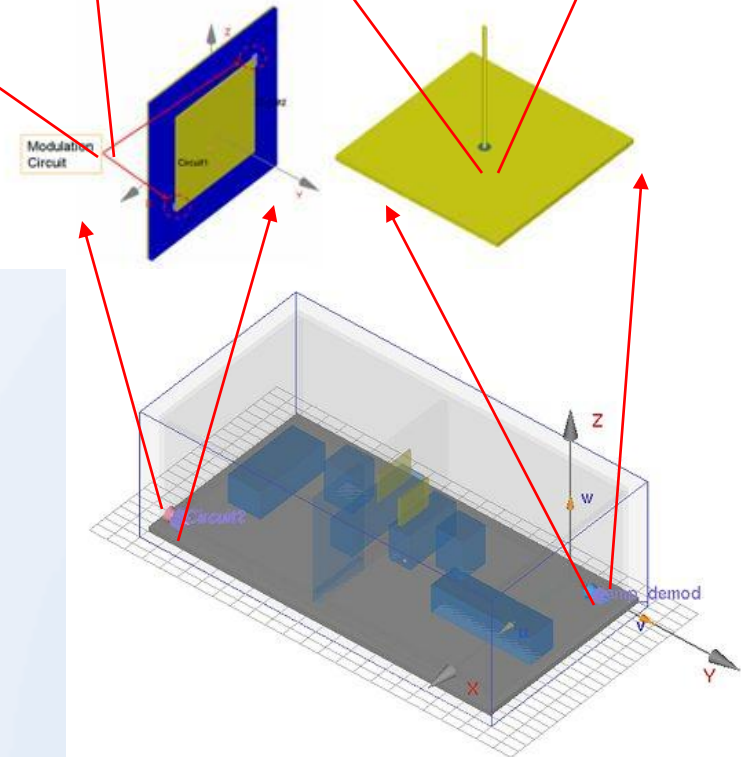
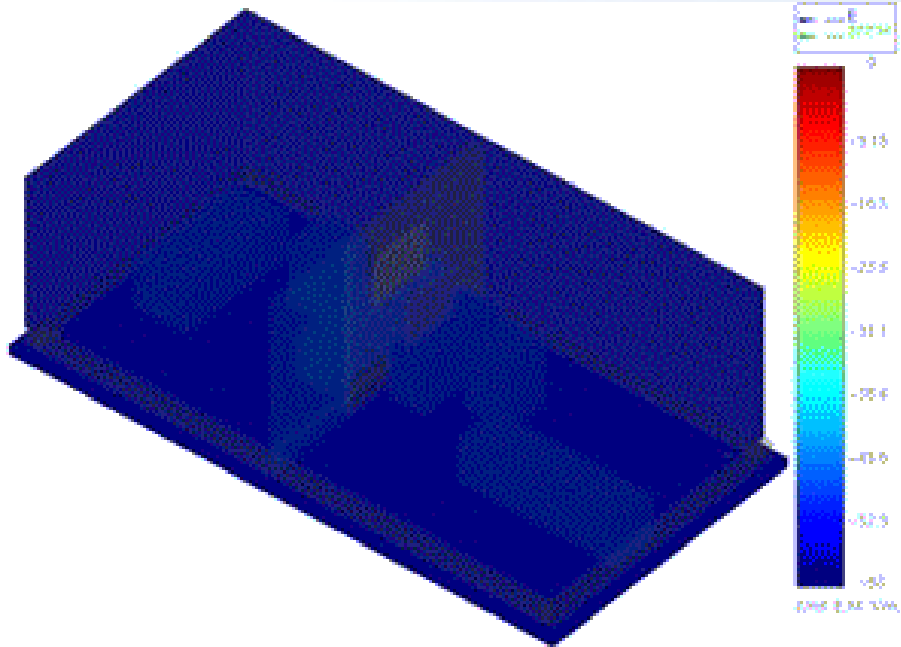
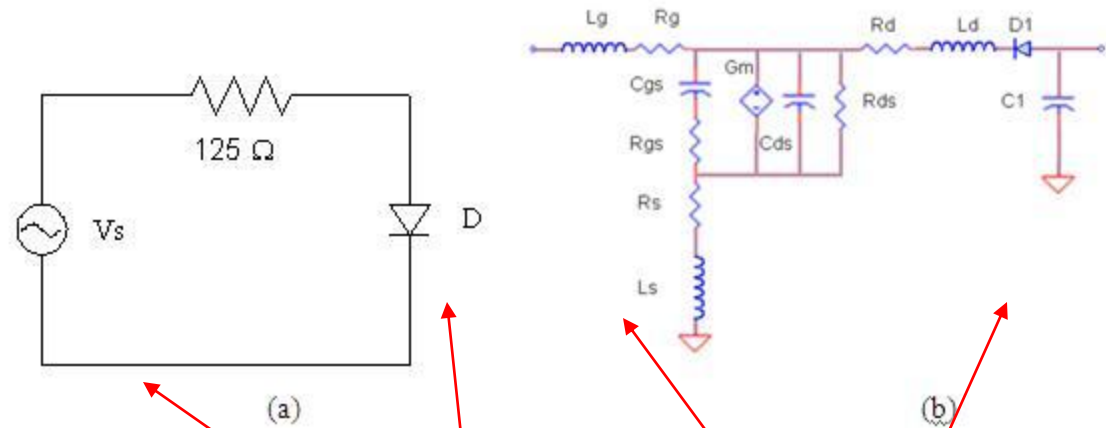


Ambient Power Collector

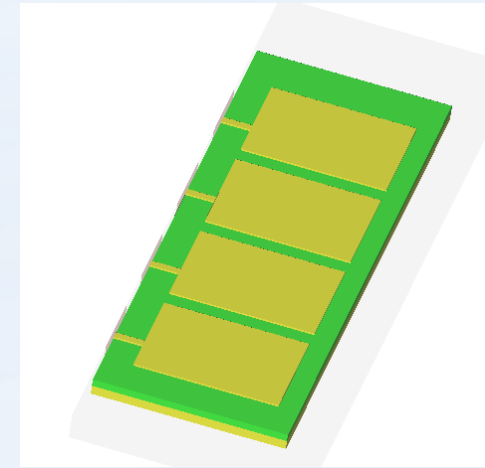
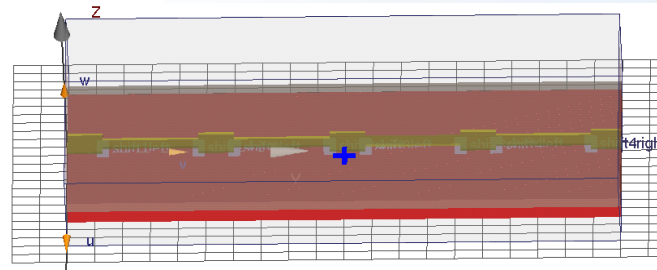
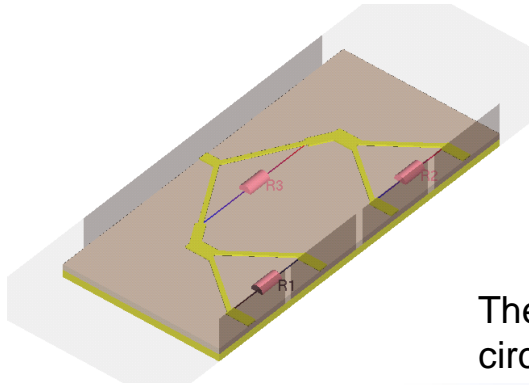


End-to-End Simulation of Antenna Systems

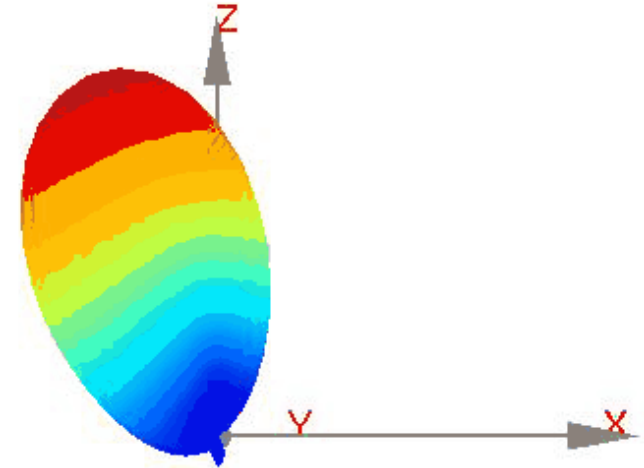
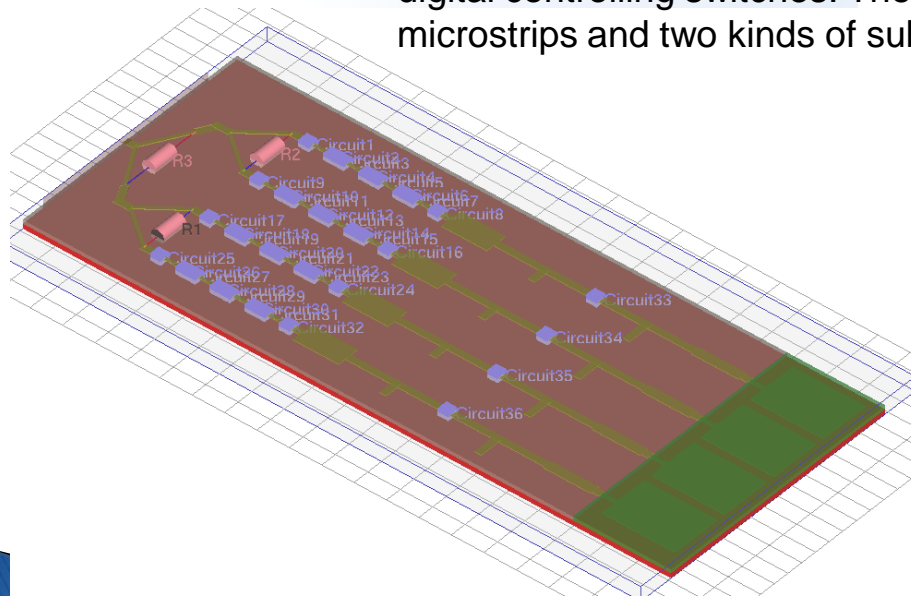
- ▶ Modulating Circuit applied on patch antenna
- ▶ Demodulating and amplification circuits applied on monopole antenna
- ▶ Two devices put in two corners of a two-bedroom model with walls, frames and furniture considered.



Solution for Phased Array Design



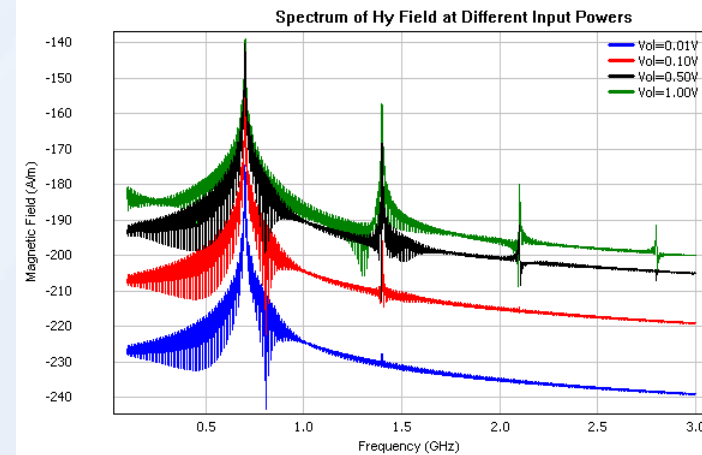
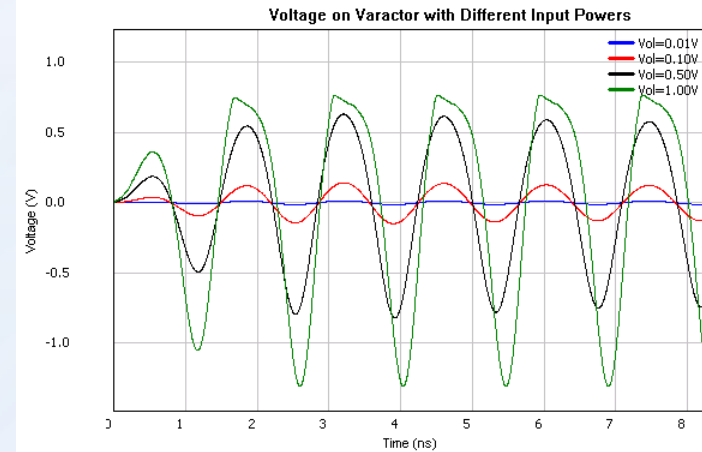
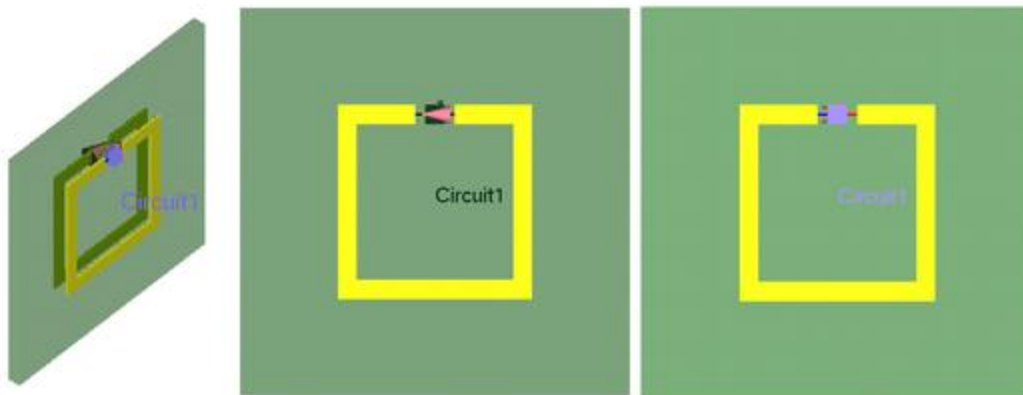
The system contains 32 independent circuits. Each circuit contains 4 capacitors, 2 resistors and two digital controlling switches. The system contains 53 microstrips and two kinds of substrates



Full-Wave Non-Linear Analysis

▶ Parametric Upconverter

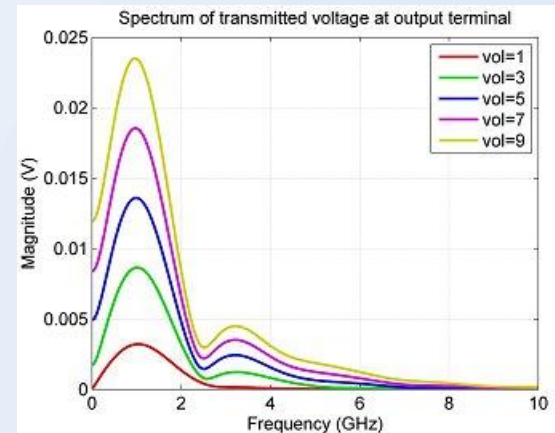
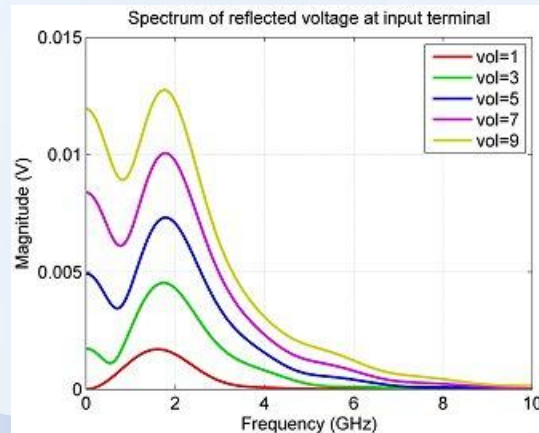
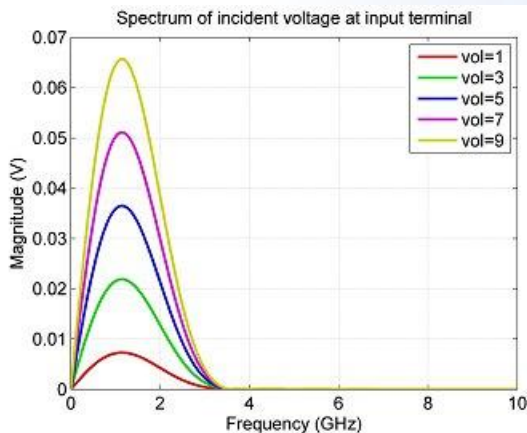
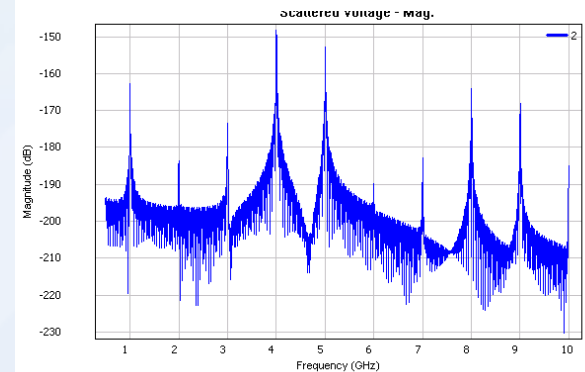
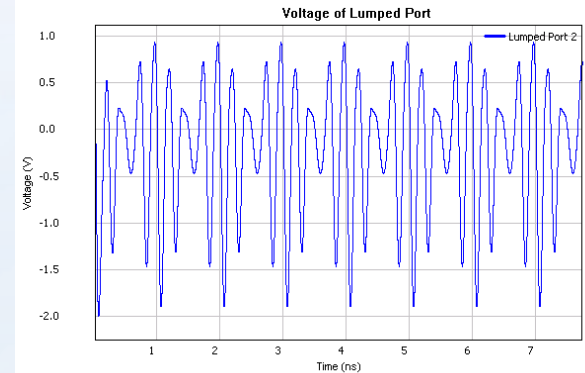
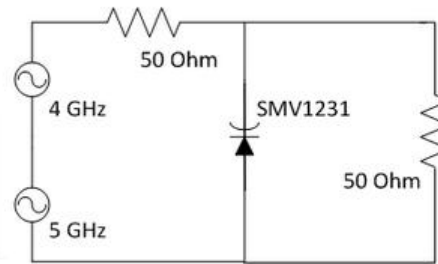
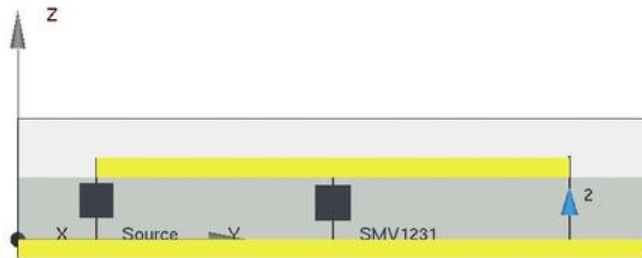
- 50 Ohm lumped port
- Varactor diode SMV1231
- 0.7GHz sinusoid source signal
- Power sweeping
 - 0.01, 0.1, 0.5 and 1V.
- Cut-off effect in transient signal
- Harmonics in spectrum



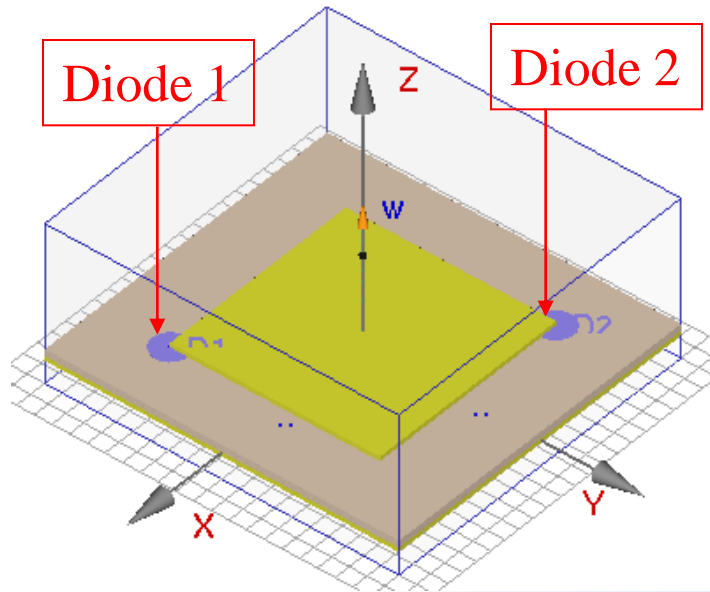
Beyond X-Parameters

▶ Parametric Amplifier

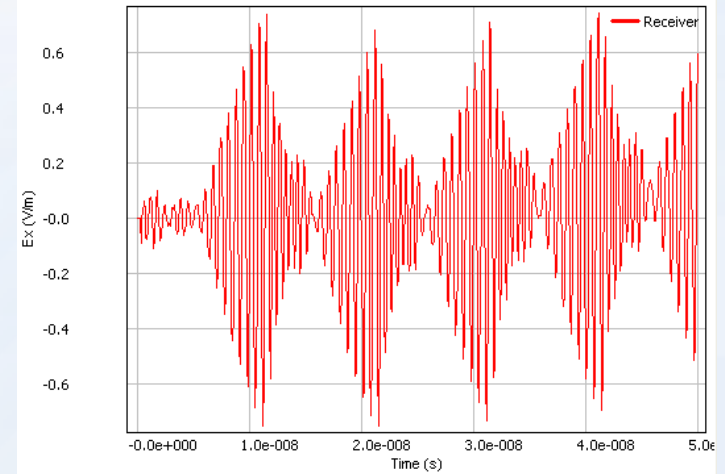
- 4GHz, 5GHz sinusoid signals are mixed.
- $f_{down}=1\text{GHz}$, $f_{up} = 9\text{GHz}$
- Wideband BHW signal also applied. Non-linear effects completely expanded the spectrum outside the input spectrum.



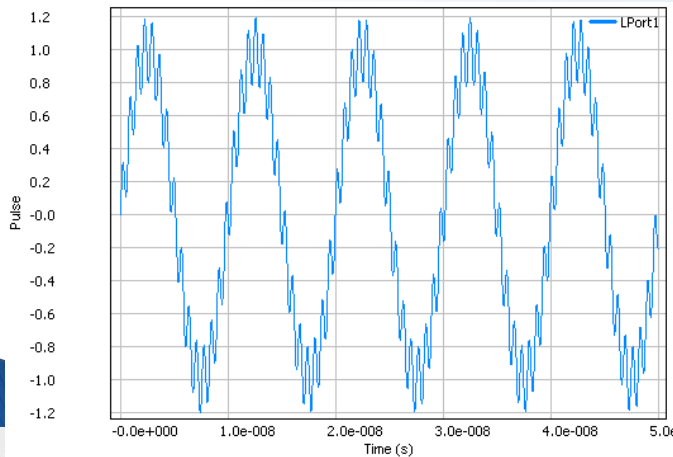
Antenna Direct Modulation



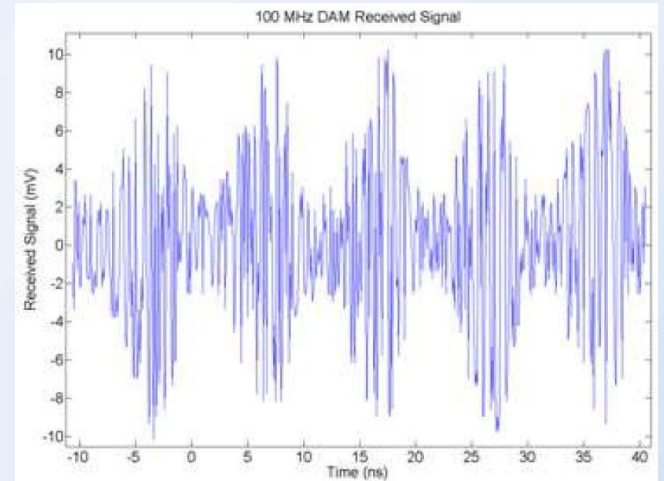
Output at receiver



Modulated input

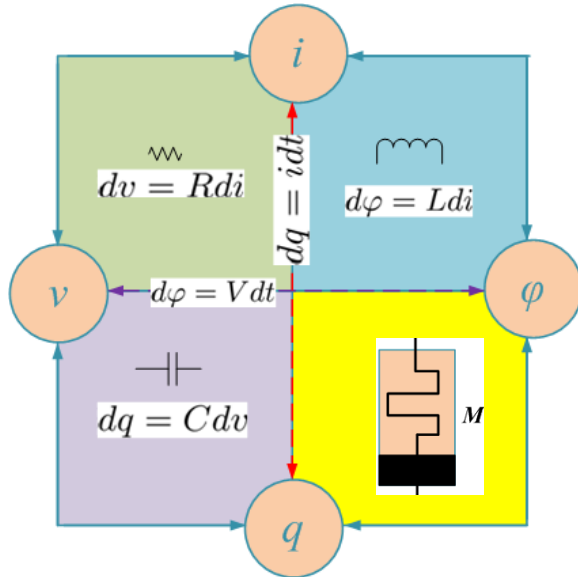


Measured



Steven David Keller, "DESIGN AND DEVELOPMENT OF DIRECTLY-MODULATED ANTENNAS USING HIGH-SPEED SWITCHING DEVICES," Thesis, Dept. of ECE, Duke Univ.

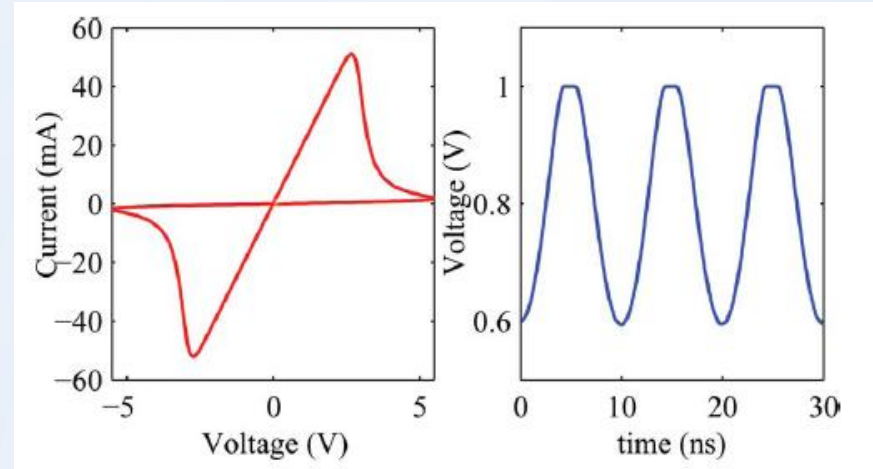
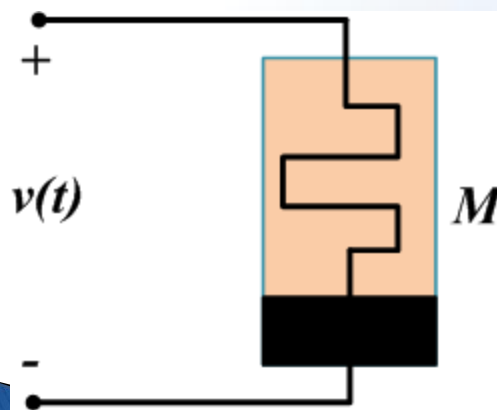
Memristor (M) – Memory Resistor: The Missing Fourth Passive Circuit Element



- 4 basic variables: v, i, q, ϕ
- 3 electrical circuit elements: R, L, C
- 2 time relationships: $v - \phi$ and $q - I t$

$$M(q(t)) = \frac{d\phi}{dq} = \frac{d\phi/dt}{dq/dt} = \frac{V(t)}{I(t)}$$

Invented by Leon Chua (1971). Produced by HP (2008)



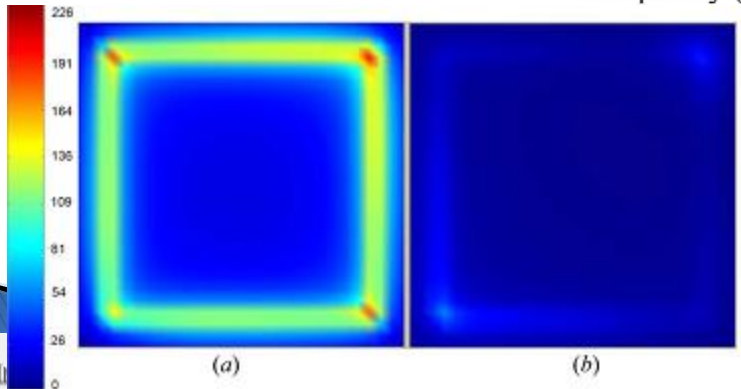
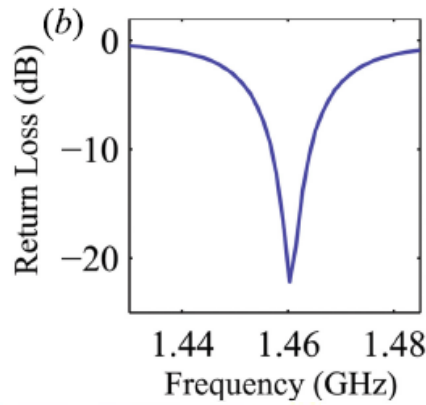
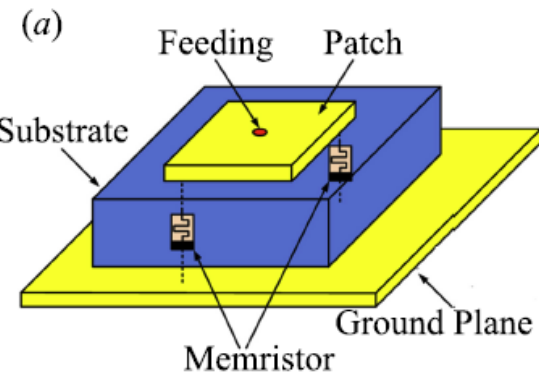
I-V hysteresis

Lin Wang; Mengqing Yuan; Tian Xiao; Joines, W.T.; Liu, Q.H.; , "Broadband Electromagnetic Radiation Modulated by Dual Memristors," *Antennas and Wireless Propagation Letters, IEEE*, vol.10, no., pp.623-626, 2011

Antenna Direct Modulation with Memristors

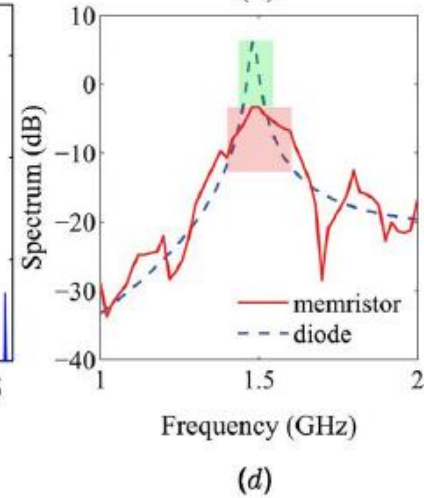
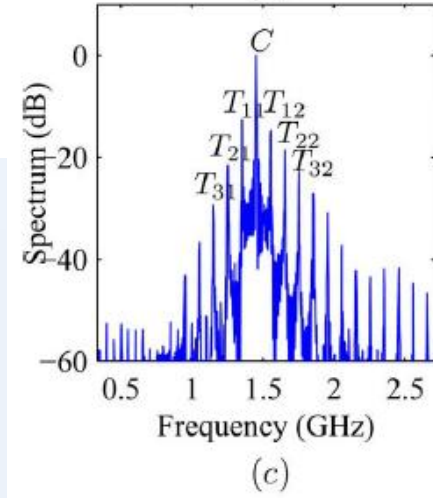
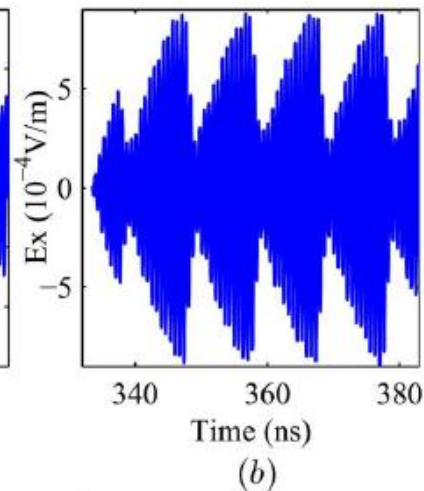
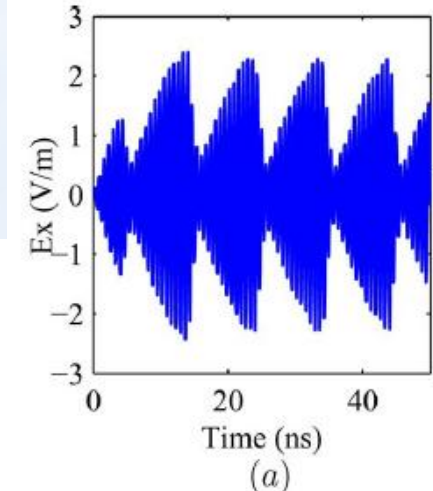
Excitation source:

- 200 MHz – 4895 MHz,
- Char. frequency: 1460 MHz

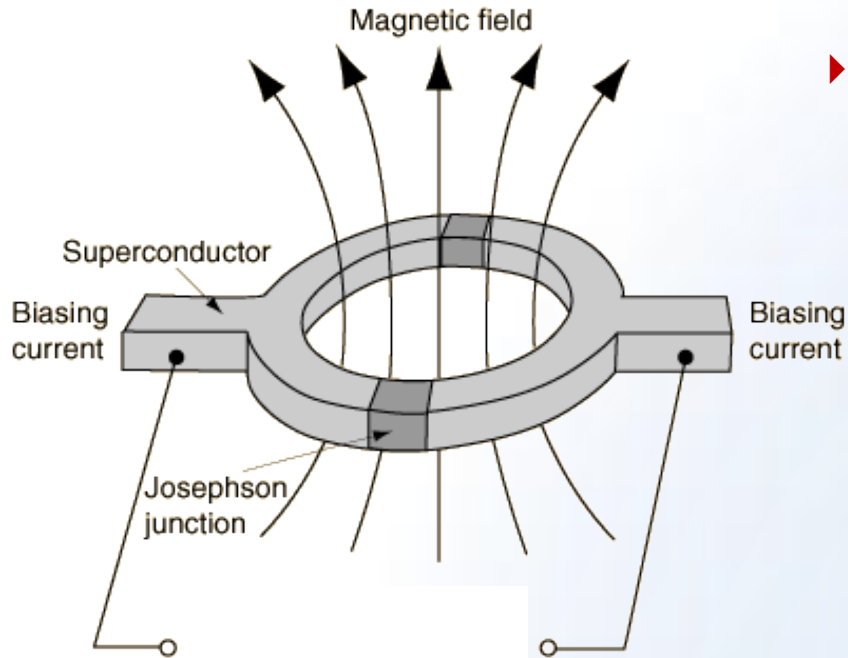


Near Field

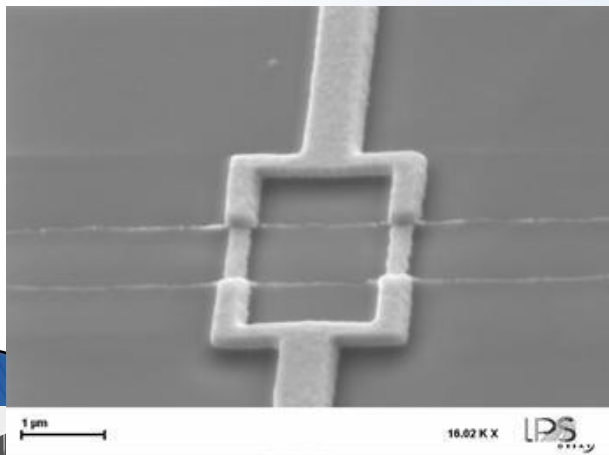
Far Field



Superconducting Devices

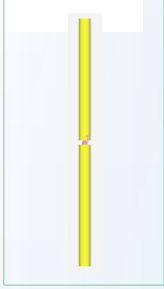
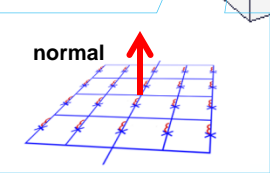
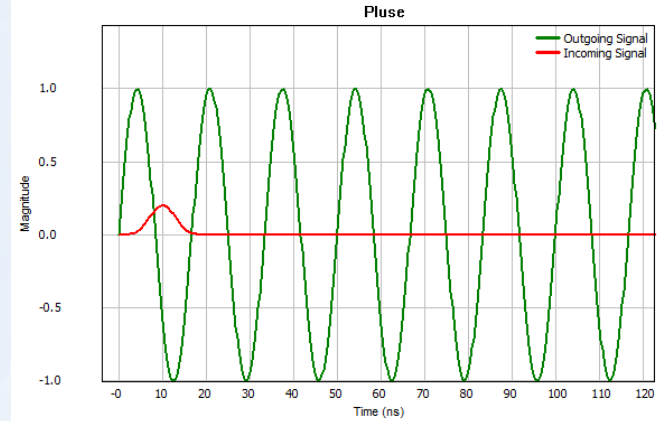
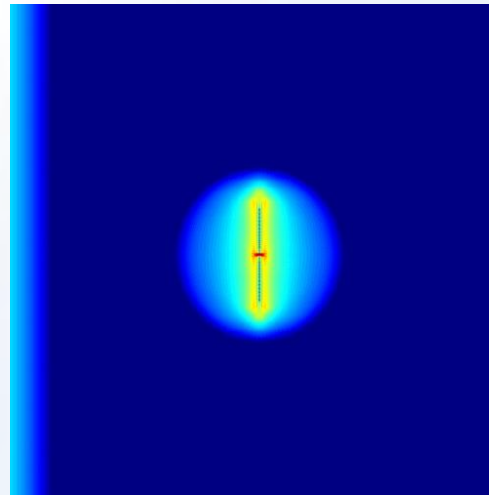
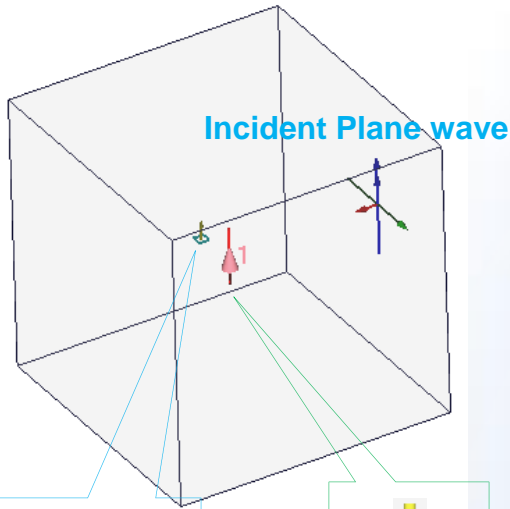


- ▶ **Superconducting Quantum Interference Devices (SQUIDs)**
 - Josephson Junctions (50–100GHz)
 - Extremely sensitive magnetic field sensors.
 - Low Noise SQUID Array Amplifiers
 - Ultra-High Resolution SQUID Magnetometers
 - SQUID Sensors for Low Frequency Imaging Applications
 - SQUID Particle and X-ray Detectors
 - SQUID Cryogenic Detector Arrays
 - SQUID Digital Processors
 - B-Field Receiving Antennas



Field-SQUID Co-Simulation

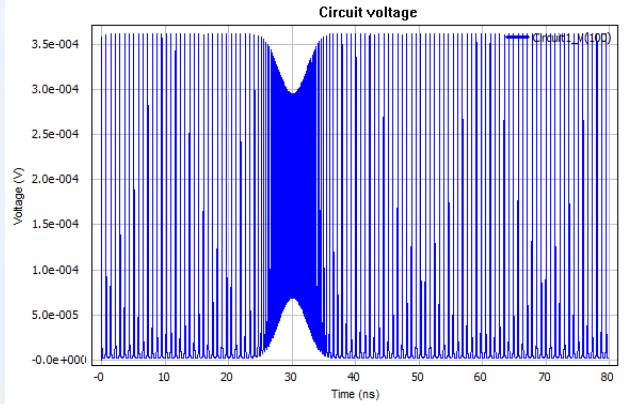
E-Transmitter: Dipole antenna
 B-Receiver: SQIF array positioned near the dipole antenna
 Transmitting signal: Sinusoid
 Receiving signal: Gaussian



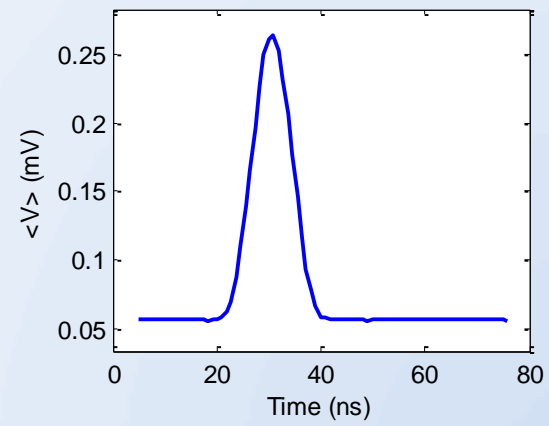
20x20 loops 2D
SQIF (B antenna)

1/2 dipole
E antenna

Transient voltage on B antenna



Averaged Voltage Output on SQIF



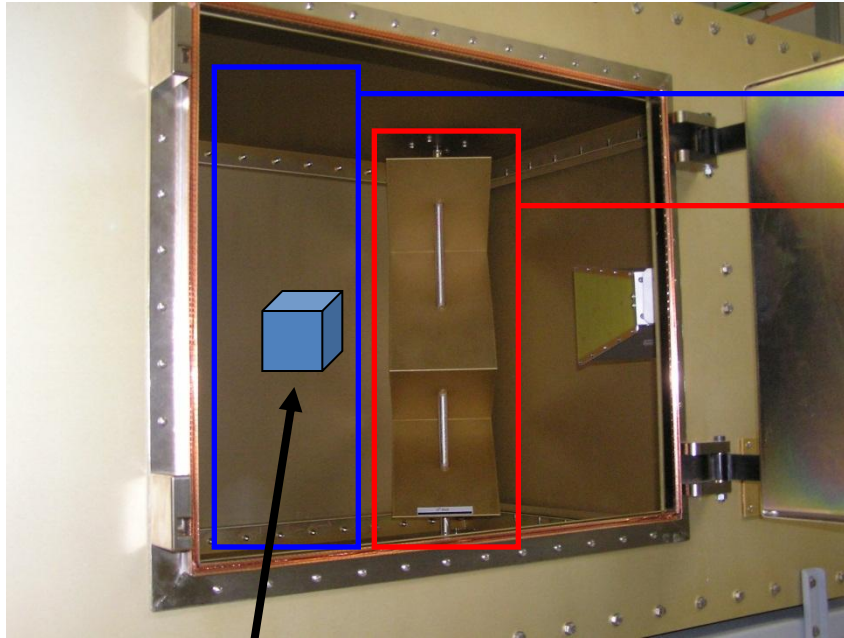
Summary

- ▶ **Field–Circuit Co–Simulation**
 - Internal Simple Circuit Elements
 - Full–Wave SPICE Analysis
 - Equivalent circuit modeling is still powerful
 - Semiconductor–based devices
 - Superconductor–based devices
 - Other novel devices
 - Non–linear effects considered in full–wave analysis
 - Coupling with EM fields is more critical as frequency goes higher and higher

Multiscale Simulation Techniques

- » Multiscale Concept
- » Review of Challenges and Solutions
- » Applications

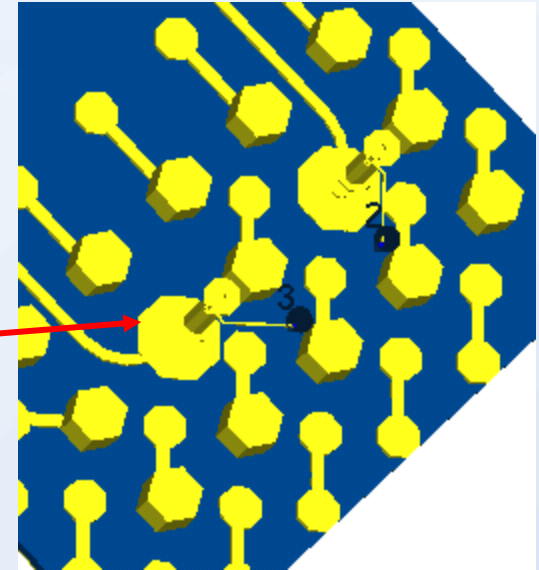
Challenge in System-Level RF Design: EMI Reverberation Chamber at 30 GHz ($\lambda=1$ cm)



- Electrically coarse structure
Chamber 1.6 m X 1.2 m X 0.8 m
- **Electrically fine structure**

DUT

Interconnect feature < 0.1 mm



No solvers can simulate such a problem on a workstation

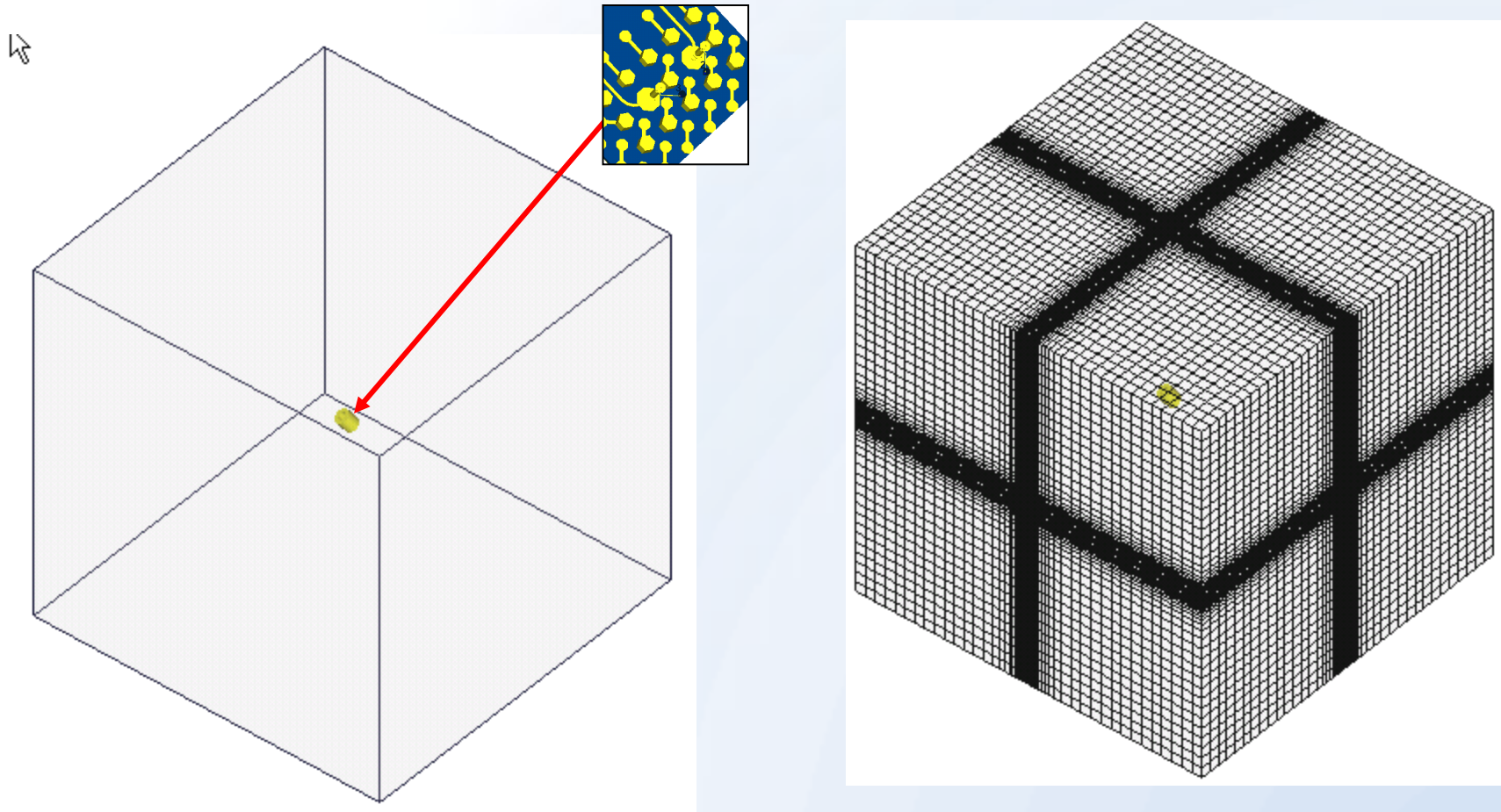
Uniform FDTD would require $40000 \times 30000 \times 20000 = 24$ trillion cells

Multiscale factor: $\frac{\Delta}{\delta} = \frac{1.6}{0.0001} = 16000$

FDTD cell: $\Delta x = 0.04$ mm

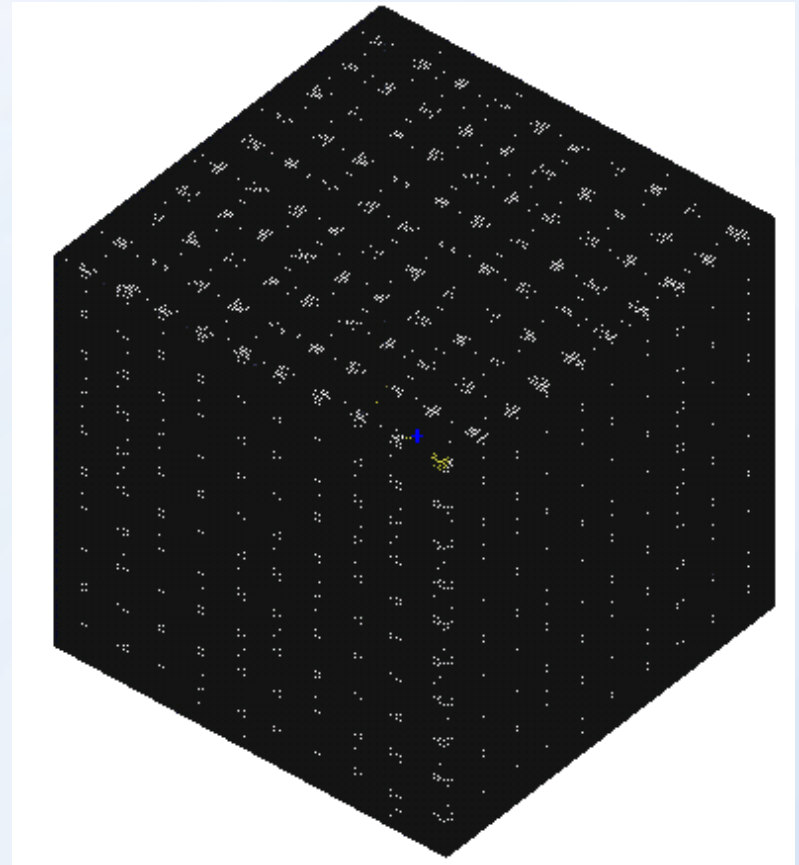
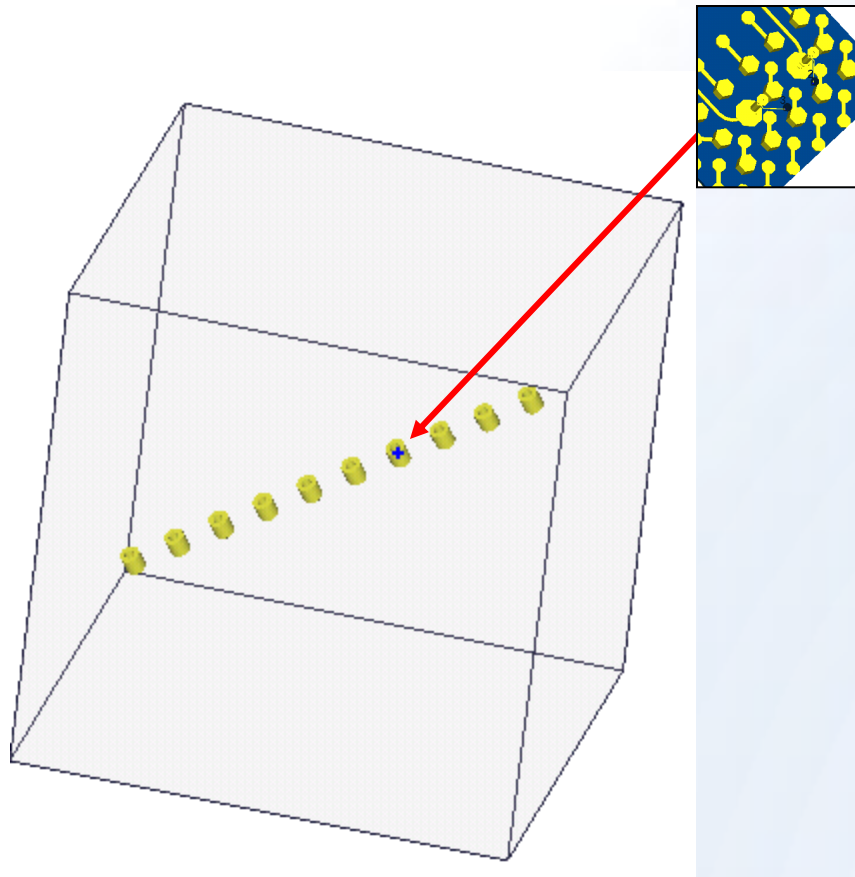
Best and Worst Scenarios for FDTD

- ▶ Best scenario: Clustered fine details



Fine cells are localized

▶ Worst scenario: Spread-out fine details



Fine cells are global

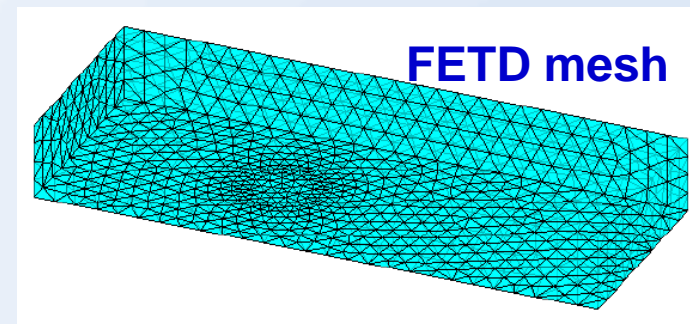
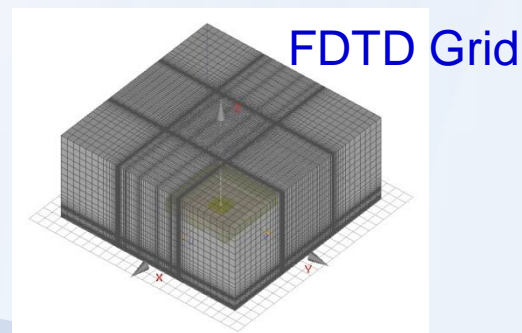
Challenges for Conventional Methods

▶ Spatial discretization

- FDTD: too many unknowns due to structured grid
- FETD: inversion or factorization of large mass matrices

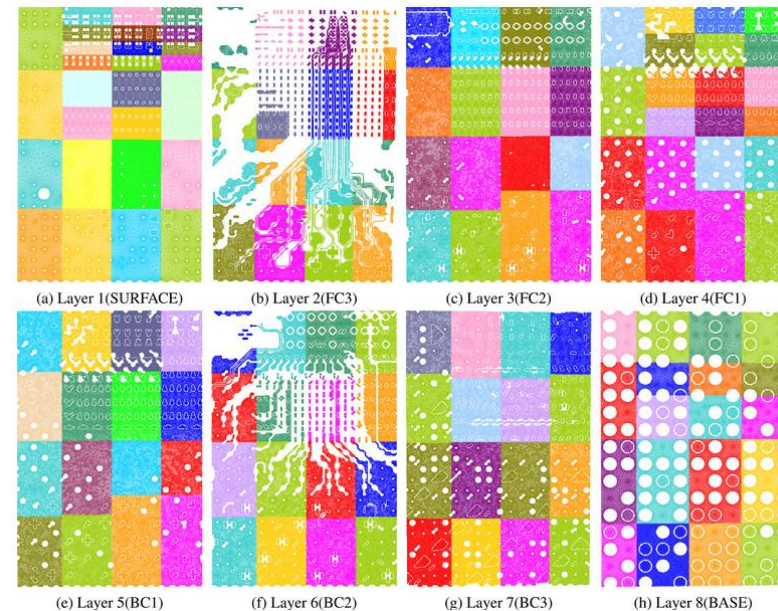
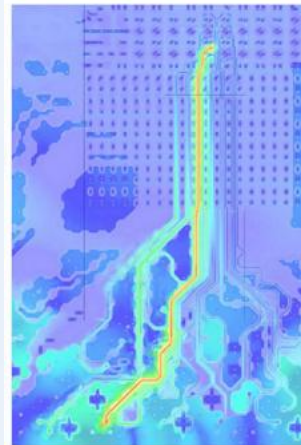
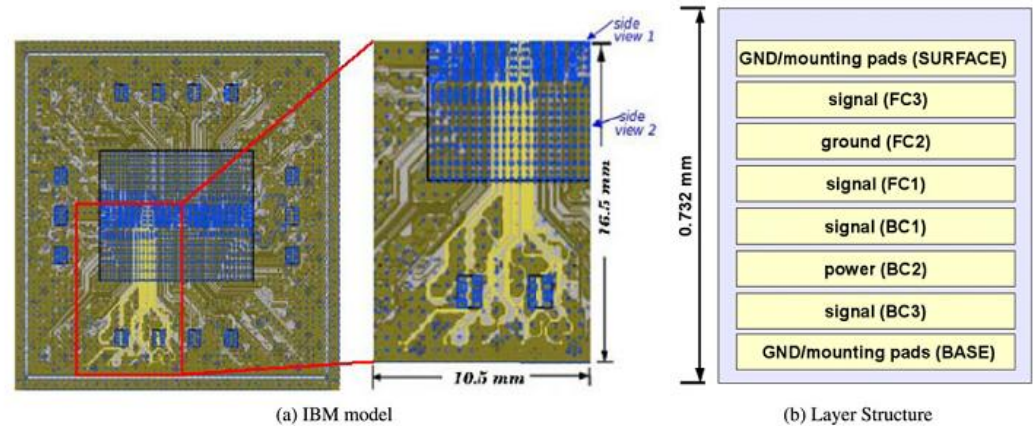
▶ Time integration

- Explicit scheme: e.g. Leap-Frog, Ex Runge Kutta
 - very small Δt due to CFL stability condition – too many time steps
- Implicit scheme: e.g. Crank-Nicolson, Im Runge Kutta
 - inversion or factorization of large matrices – large memory and CPU time



Transient DDM within Finite Element Method

- ▶ **Flux Operation**
 - Central Flux
 - Upwind Flux
- ▶ **Domain Interface**
 - Conformal
 - Non-Conformal
- ▶ **Parallel Computing**
 - MPI/GPU/Multithreading
 - Load Balancing



Stylianos Dosopoulos, Bo Zhao, Jin-Fa Lee, Non-conformal and parallel discontinuous Galerkin time domain method for Maxwell's equations: EM analysis of IC packages, Journal of Computational Physics, Volume 238, 1 April 2013, Pages 48-70

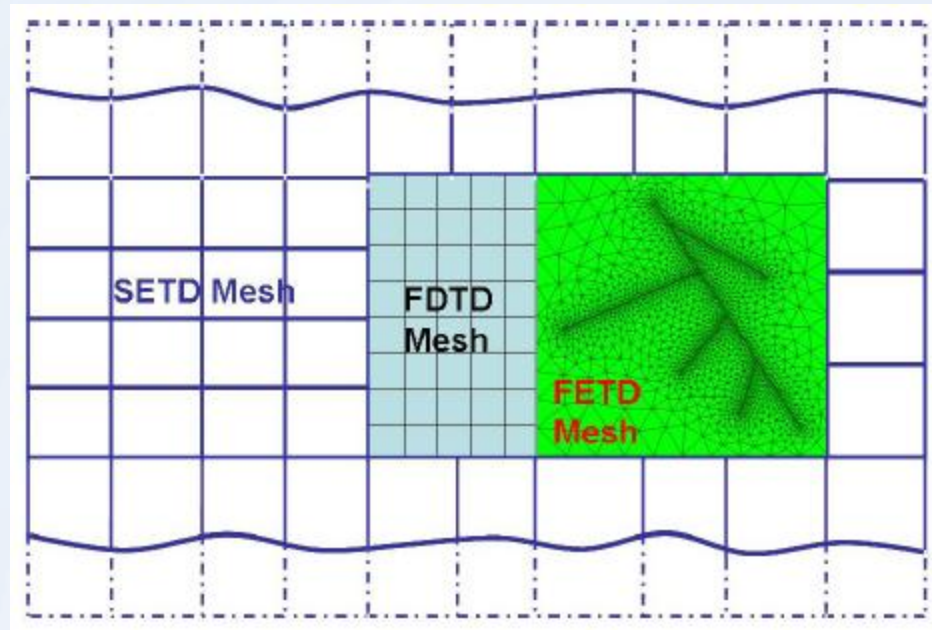
Multiscale Hybrid SETD/FDTD/FETD Method

- ▶ Electrically fine structures: lower order tetrahedral FETD
- ▶ Electrically coarse structures: higher order hexahedral SETD
- ▶ Intermediate structures: boundary conformal FDTD
- ▶ Interface between different subdomains: Riemann solver

Riemann Solver for interface

$$\begin{aligned} (Y^{(i)} + Y^{(j)}) (\mathbf{n} \times \mathbf{E}) &= \mathbf{n} \times (Y^{(i)} \mathbf{E}^{(i)} + Y^{(j)} \mathbf{E}^{(j)}) \\ &\quad - \mathbf{n} \times \mathbf{n} \times (\mathbf{H}^{(i)} - \mathbf{H}^{(j)}) \end{aligned}$$

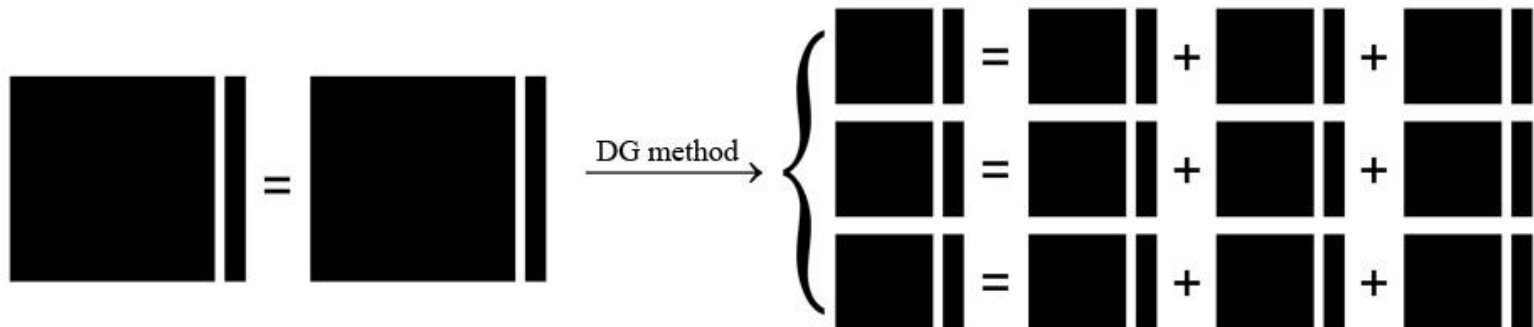
$$\begin{aligned} (Z^{(i)} + Z^{(j)}) (\mathbf{n} \times \mathbf{H}) &= \mathbf{n} \times (Z^{(i)} \mathbf{H}^{(i)} + Z^{(j)} \mathbf{H}^{(j)}) \\ &\quad + \mathbf{n} \times \mathbf{n} \times (\mathbf{E}^{(i)} - \mathbf{E}^{(j)}) \end{aligned}$$



Local adaptive time integration scheme: Implicit-explicit Runge-Kutta (IMEX-RK)

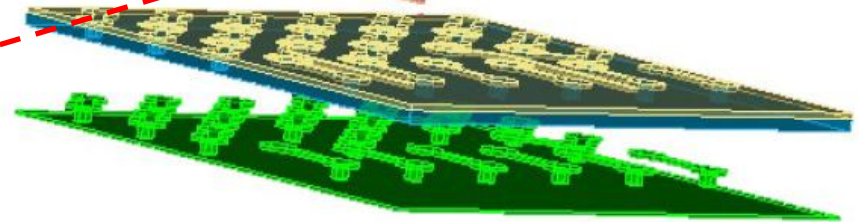
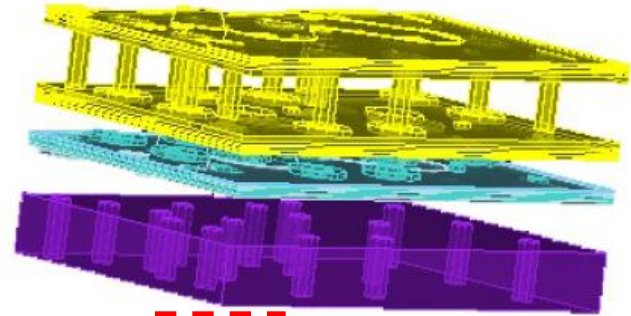
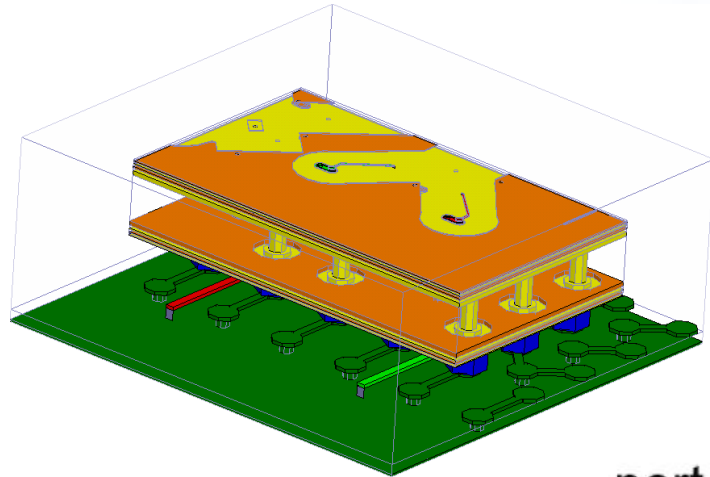
- ▶ Electrically coarse subdomains: explicit Runge-Kutta scheme
- ▶ Electrically fine subdomains: implicit Runge-Kutta scheme
- ▶ Adjacent explicit and implicit subdomains: IMEX-RK scheme

$$\mathbf{v}^{(i)}(t_{n+1}) = \mathbf{v}^{(i)}(t_n) + \Delta t \sum_{k=1}^s \sum_{j=1}^{N_{\text{ex}}} b_k^{\text{ex}} \mathbf{T}_{\text{ex}}^{(ij)} \mathbf{u}_{\text{ex}}^{(j)}(k) + \Delta t \sum_{k=1}^s \sum_{j=1}^{N_{\text{im}}} b_k^{\text{im}} \mathbf{T}_{\text{im}}^{(ij)} \mathbf{u}_{\text{im}}^{(j)}(k)$$



Large system matrices are divided into several middle sized matrices by the hybrid method

Interconnect package



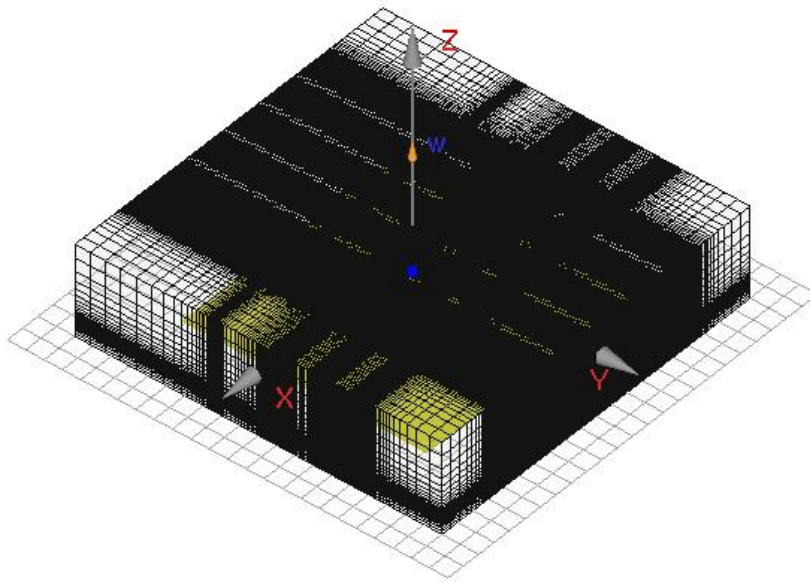
port 1



port 2

FDTD grid and SETD/FETD mesh

Jiefu Chen; Tobon, L.E.; Mei Chai; Mix, J.A.; Qing Huo Liu; , "Efficient Implicit-Explicit Time Stepping Scheme With Domain Decomposition for Multiscale Modeling of Layered Structures," *Components, Packaging and Manufacturing Technology*, *IEEE Transactions on* , vol.1, no.9, pp.1438-1446, Sept. 2011

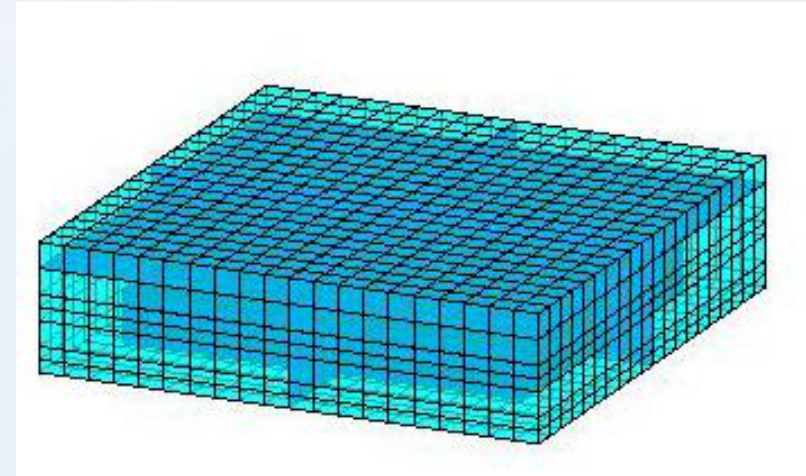


FDTD grid
PPW=40

cells: 511 X 323 X 60
total DoF: > 50 million

$\Delta t = 3.98$ fs

nt = 125,628



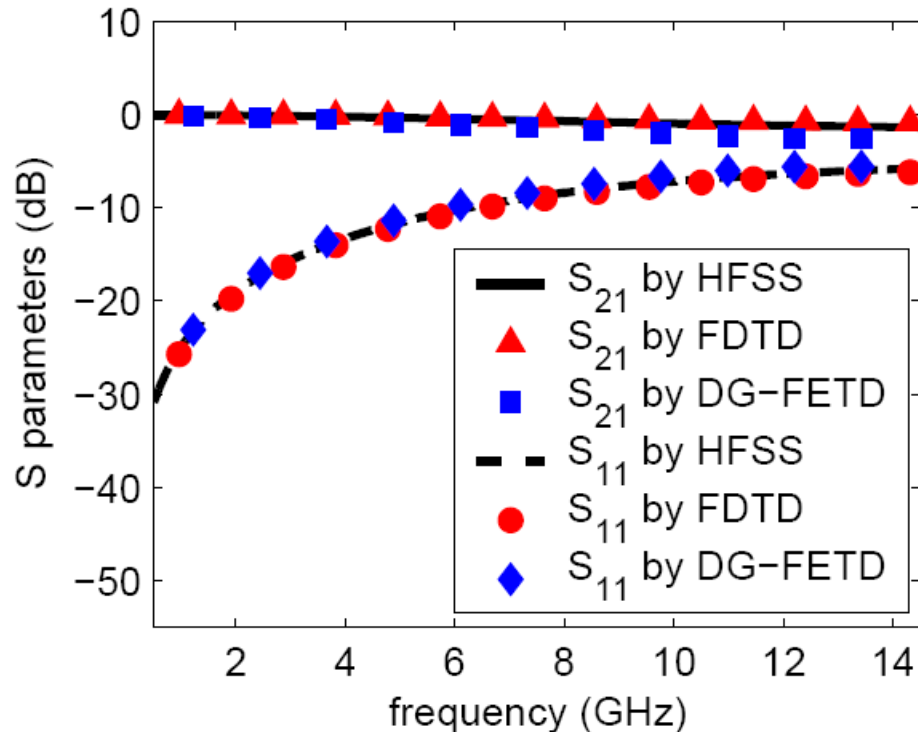
SETD / FETD mesh
PPW=40

44 subdomains
total DoF: 152,356

$\Delta t = 500$ fs

nt = 1,000

Numerical results by three methods

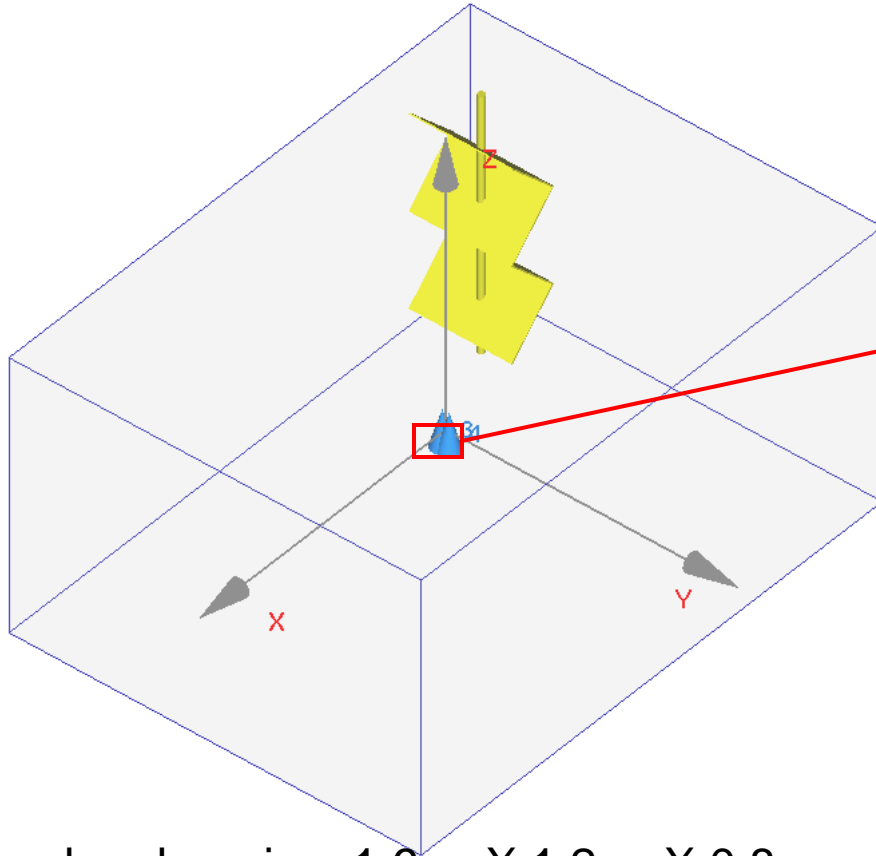


	hybrid SETD/FETD	FDTD	HFSS
memory (MB)	371	1,627	1,433
CPU time (minutes)	13.1	522	319

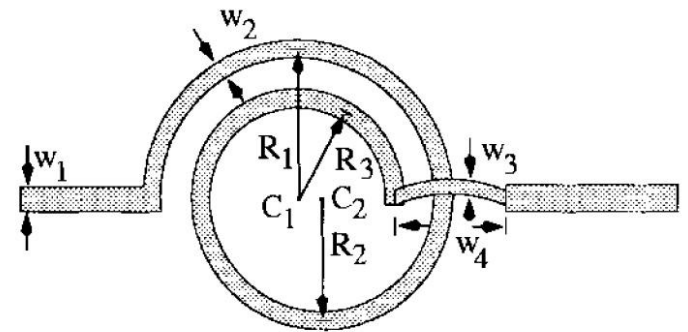
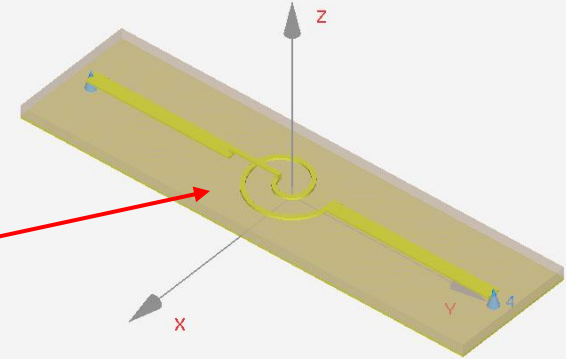
(efficiency ratio) hybrid : FDTD = 39

(efficiency ratio) hybrid : HFSS = 24

Spiral Inductor in a Reverberation Chamber (3 GHz)



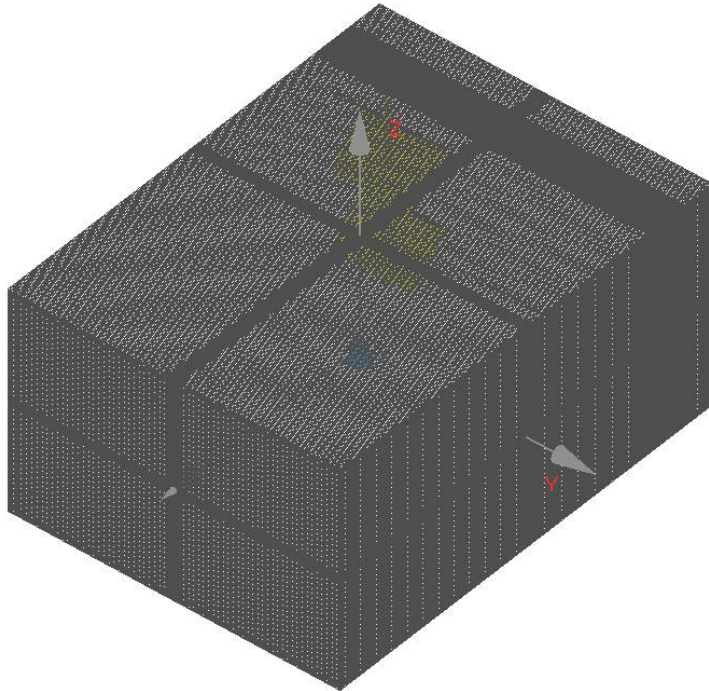
chamber size: 1.6 m X 1.2 m X 0.8 m



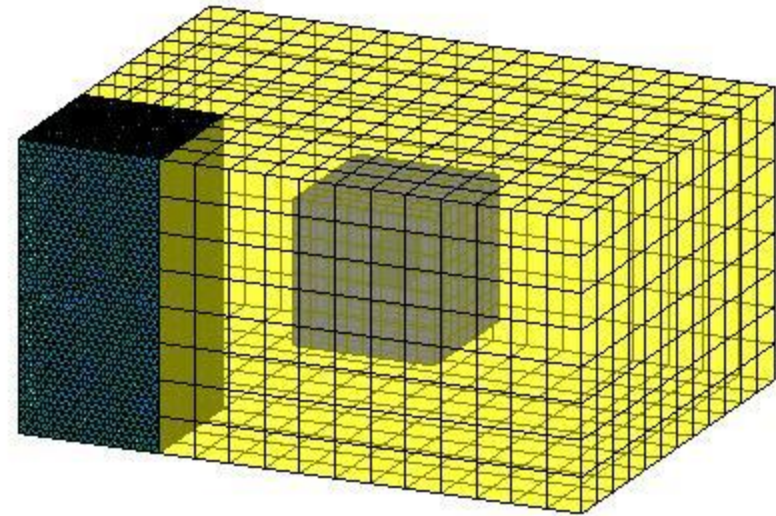
$$w_1 = 0.635 \text{ mm}, w_2 = w_3 = 0.2 \text{ mm}, w_4 = 2.4$$

$$R_1 = 1.9 \text{ mm}, R_2 = 1.3 \text{ mm}, R_3 = 0.7 \text{ mm}$$

FDTD Grid Versus SETD/FETD Mesh

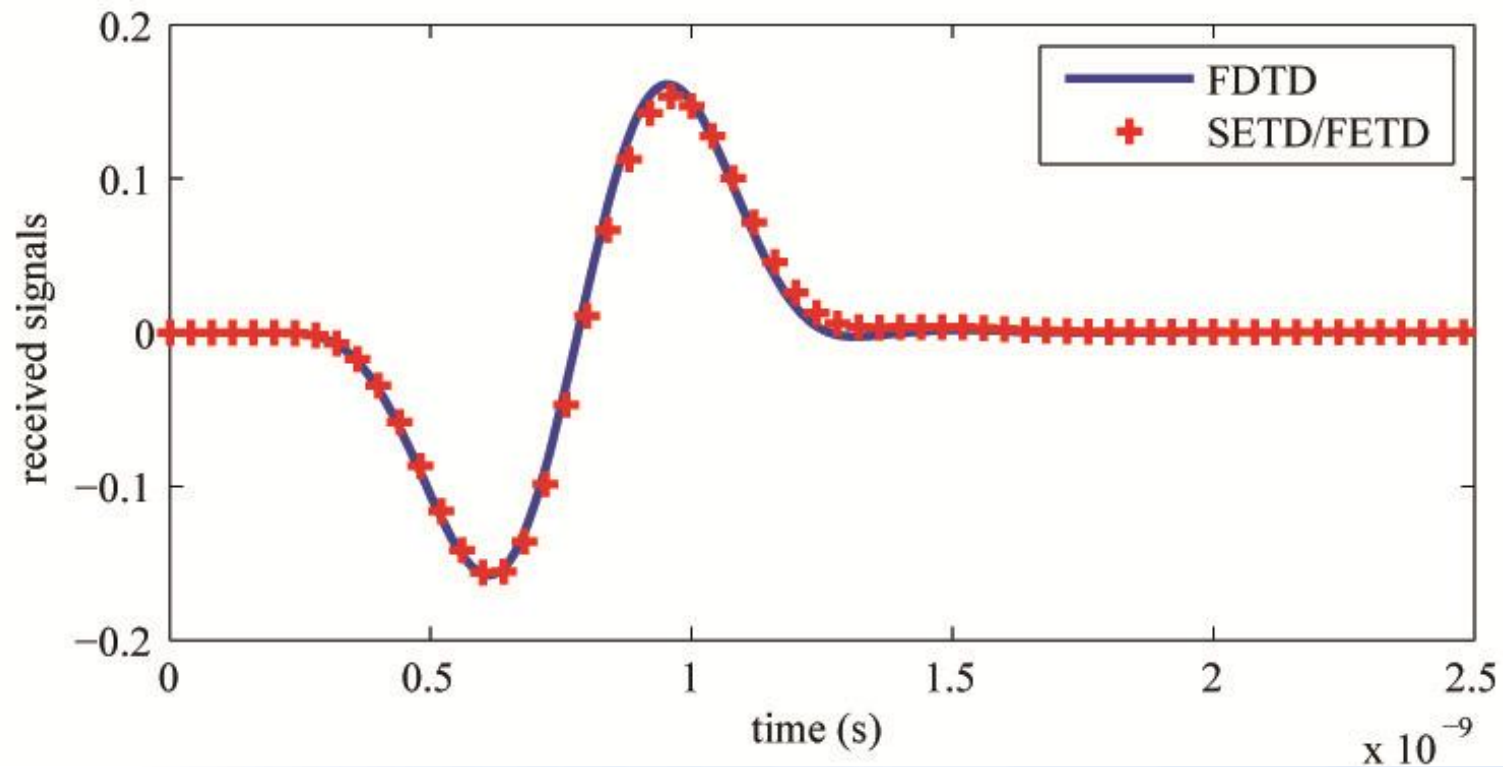


(PPW=16 with local refinement)
 FDTD grid: 473 X 420 X 167
 DoF: 199.1 million
 memory cost: 3.3 GB
 maximum $\Delta t = 0.137$ ps



(similar discretization as FDTD
 for fine structures and stirrer)
 DoF: 1,654,475
 Memory cost: 840 MB
 Δt for IMEXRK = 10 ps

relative difference between FDTD and SETD/FETD = 8.5 %

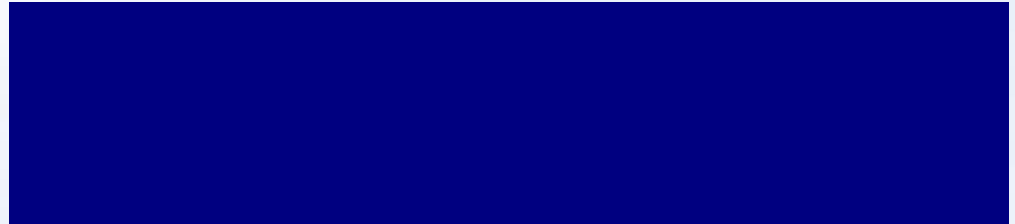


Tab. 1: comparison of computational costs

	FDTD	SETD/FETD	gain by SETD/FETD
memory (MB)	3379	840	4.0
CPU time (h)	19	7.2	2.6

EMP/EMI of Unmanned Aerial Devices/Vehicles

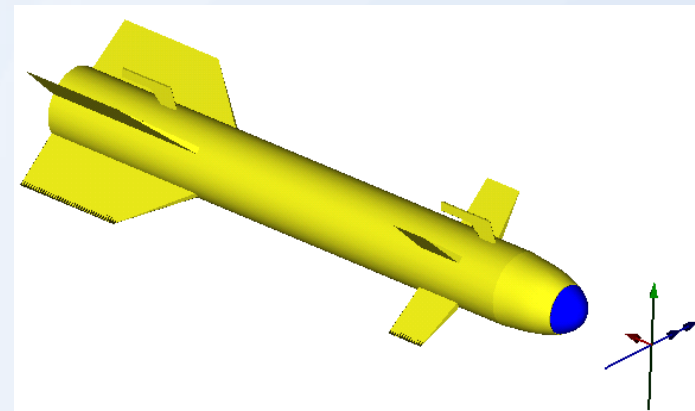
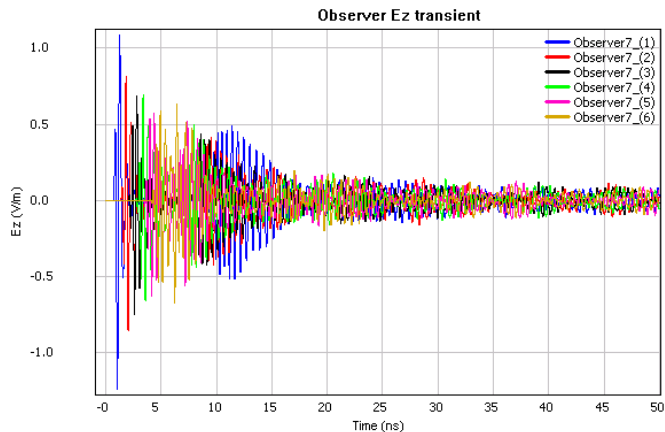
E field snapshot



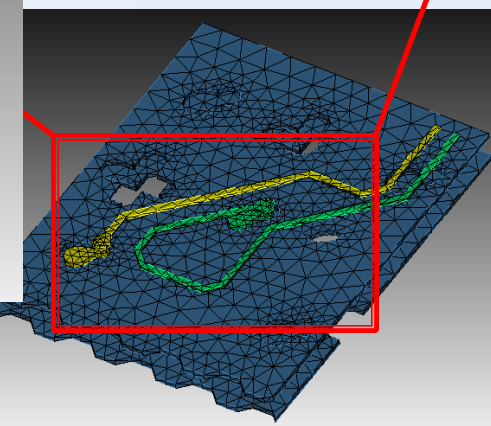
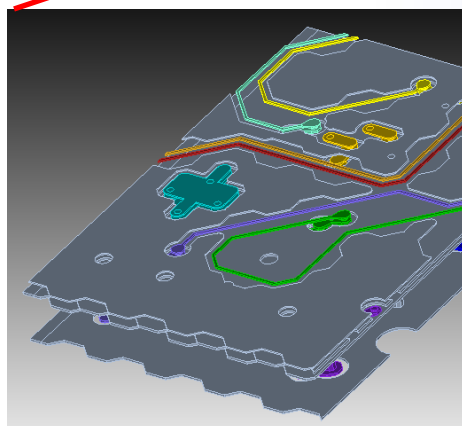
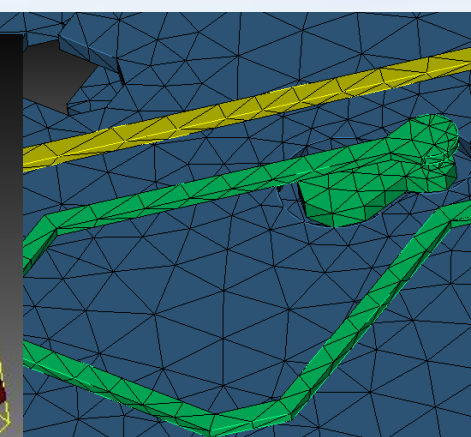
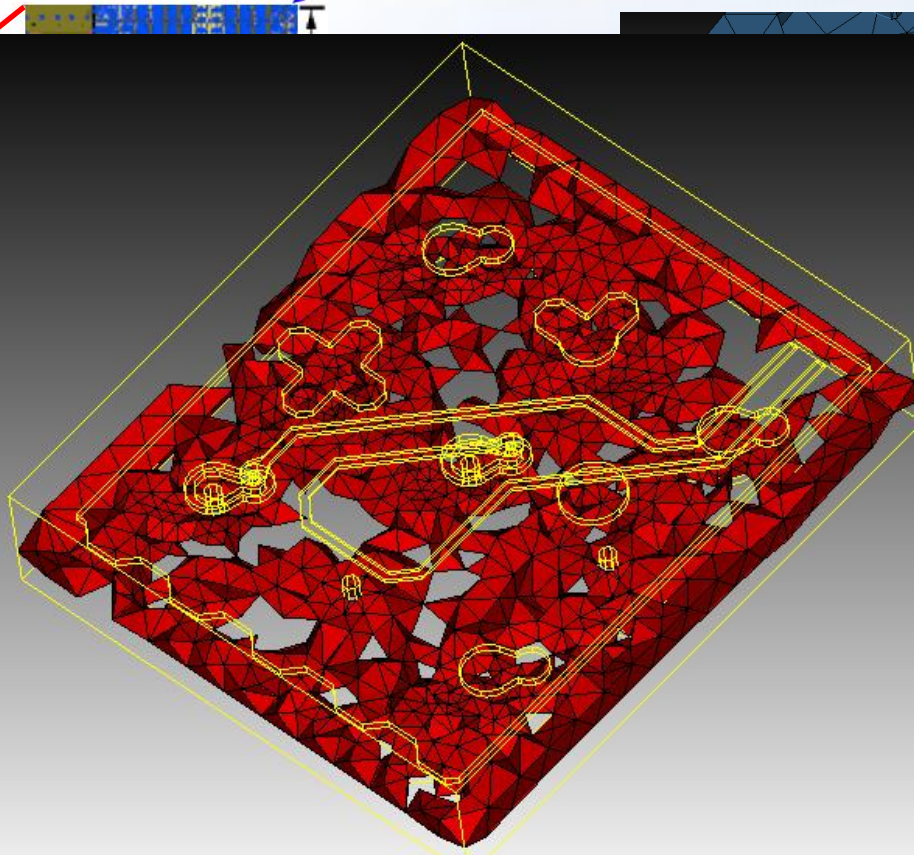
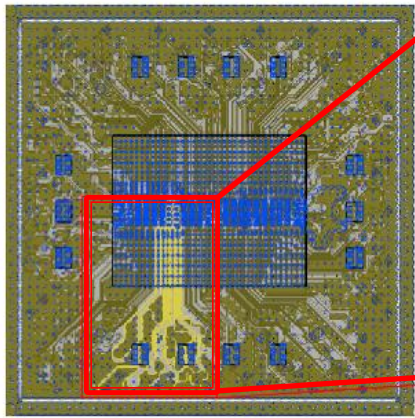
H field snapshot



HELLFIRE II

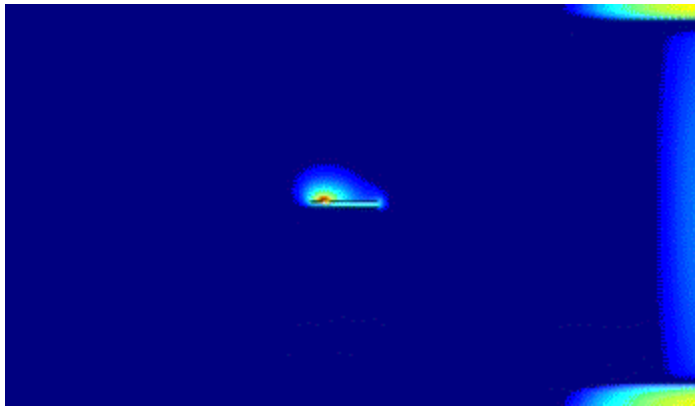


Capturing the Details

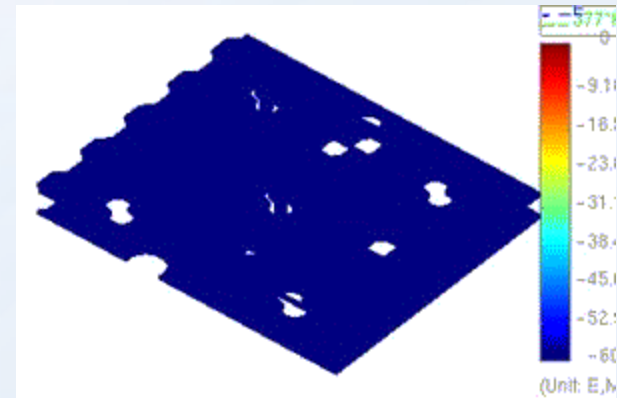


Reveal the EM Interference

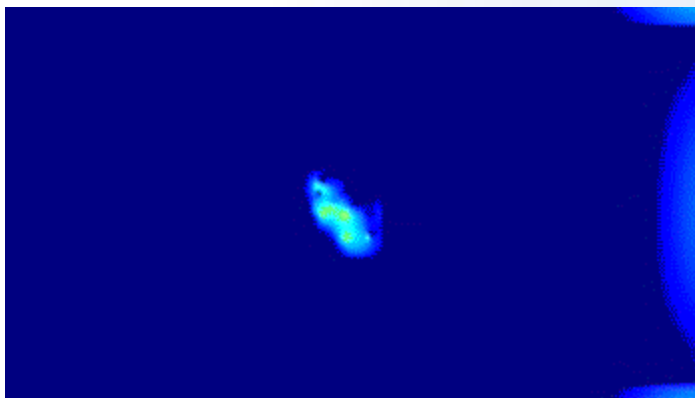
E field



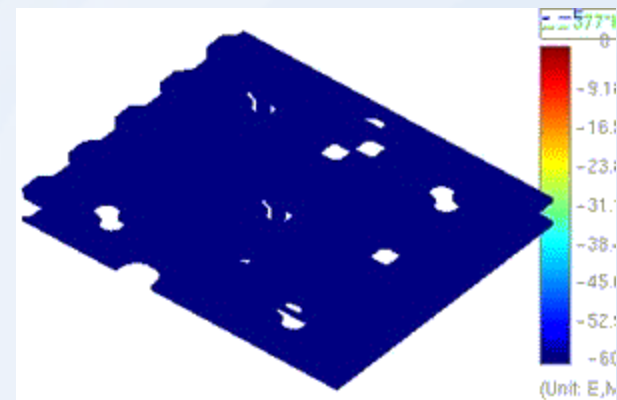
Current generated from source signal



H field



Current generated from interference signal

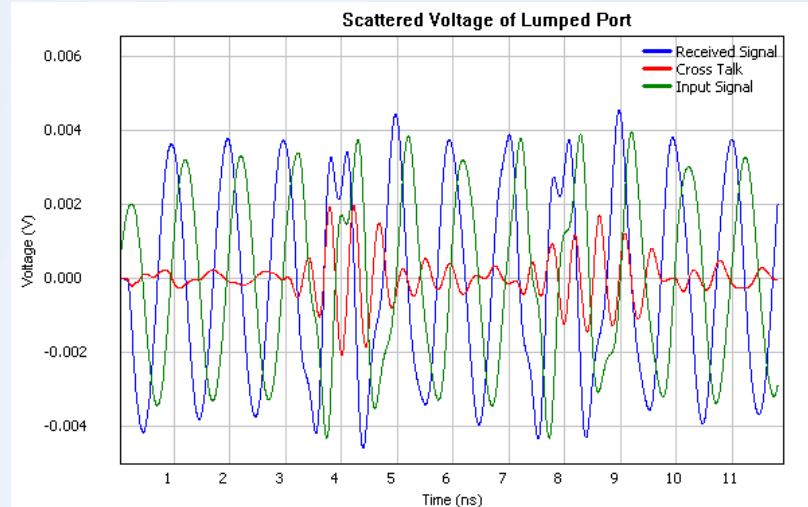
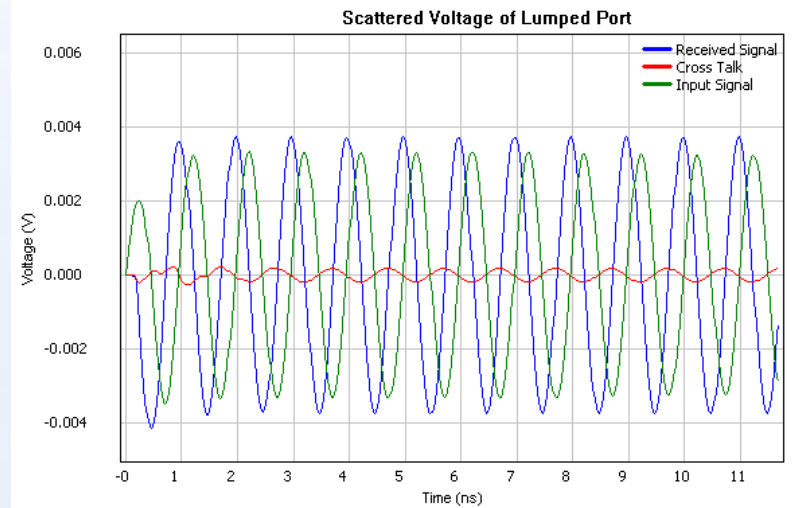
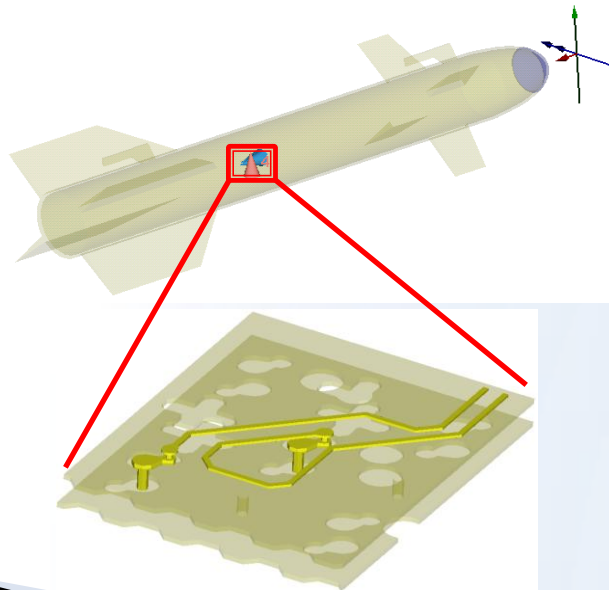


Resonating EM field interacting with the circuit board

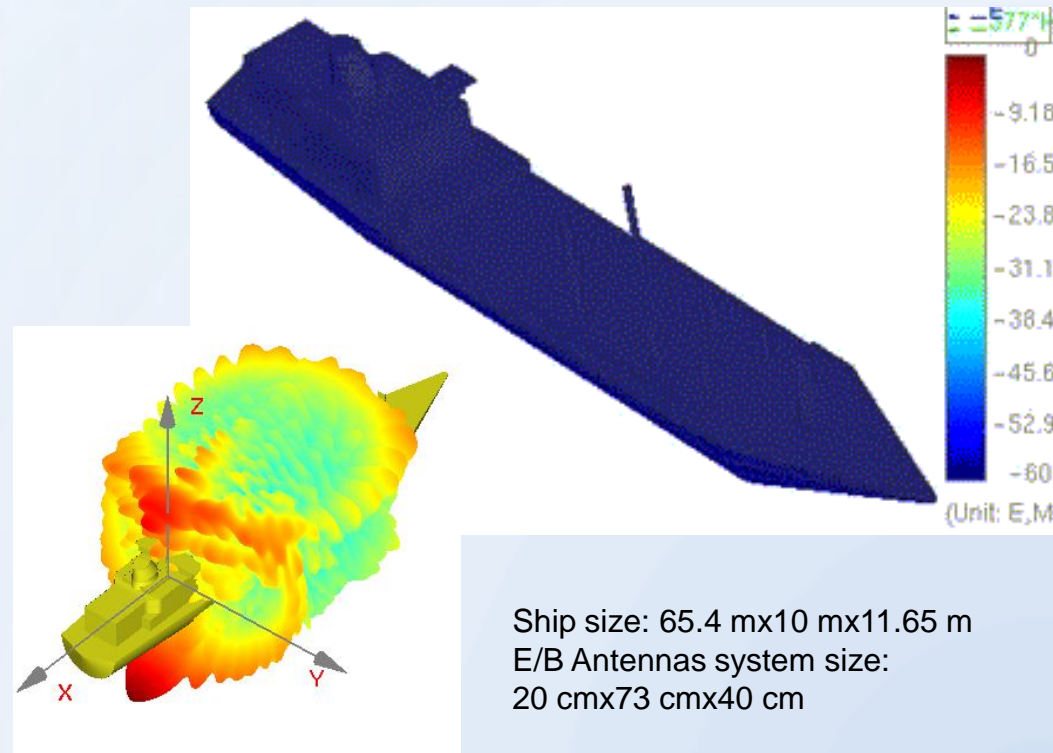
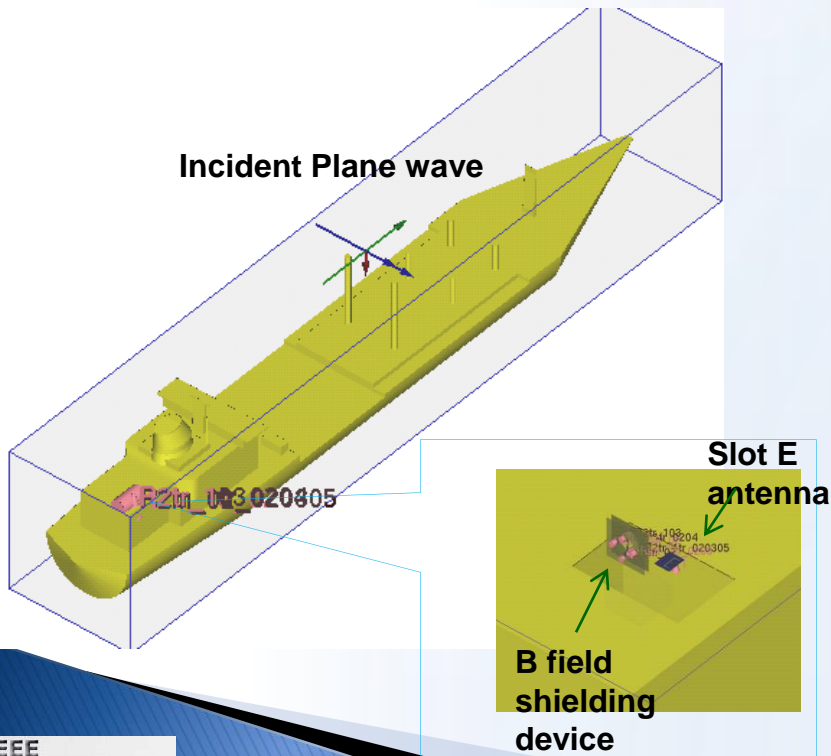
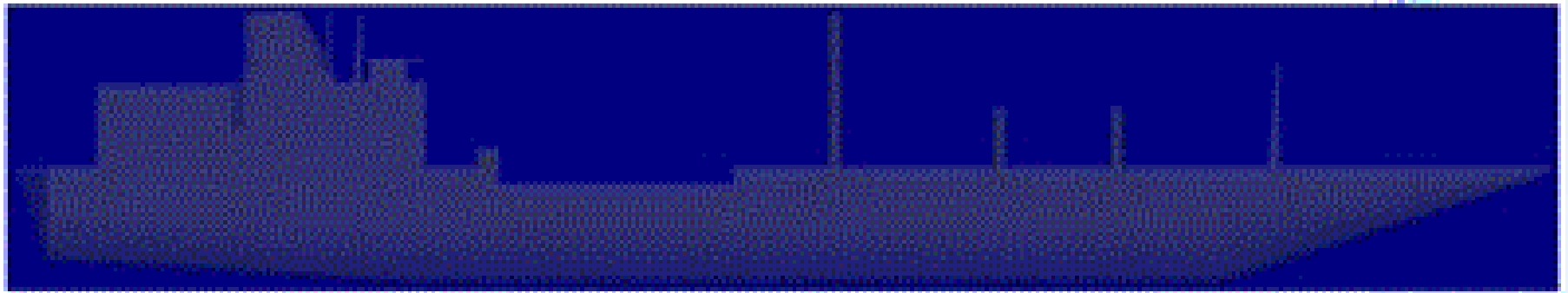
Surface current on the circuit board

Reveal the EM Interference (Cont'd)

- ▶ Circuit board inside missile cavity
- ▶ Clean signal vs. interfered signal



Other Novel Applications using Hybrid Multiscale Simulator



Summary

▶ Multiscale EM field solver

	FD	FE	SE
Efficiency	5	1	3
Accuracy	1	5	3

▶ DG is the key

- ▶ Domain decomposition
 - ▶ Geometry decomposition
 - ▶ Flux operation

▶ Adaptive Time Stepping

- ▶ LTS
- ▶ IM-EX

About the Company

- » Profile
- » Products

Company Profile

► Overview

- Type of business: Research and development
- Products: Engineering simulation software
- Company Sites: Research Triangle Area, NC
- Number of personnel: 5

► History

- Founded in 2005
- 5 Phase-I and 3 Phase-II SBIR Projects
 - Air Force (2), Navy (3), Army (2), NIH (1)
- In Process on 3 projects
 - Navy (Phase-II.5), Department of Energy (pending), NASA (pending)

▪ Reference:

<http://www.sbir.gov/sbirsearch/detail/349368>



Commercialized Products

- ▶ **Wavenology EM (electromagnetics)**
 - General purpose transient EM field simulator. CAD tool for design of smart antenna, RF/microwave circuit system and novel devices.

- ▶ **Wavenology PIC (particle in cell)**
 - Designs of EM railguns, accelerators and other particle devices

- ▶ **Wavenology EL (elastrodynamics).**
 - An advanced elastic wave simulator. It focuses on oil exploration with ultrasonic, sonic and seismic waves, with major oil services companies as our clients.

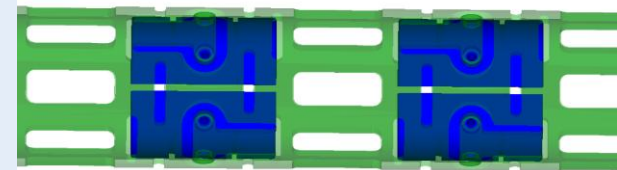
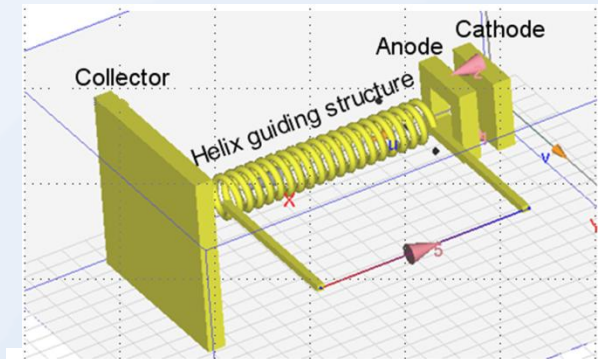
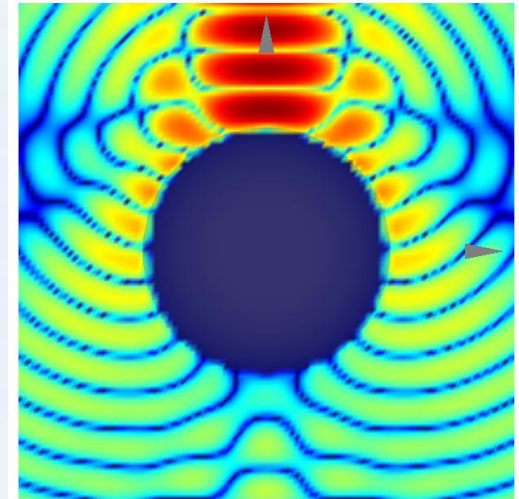


Image Citations

- ▶ Page 4: Top left figure is from <http://www.rsphysse.anu.edu.au/nonlinear/research/lhm/>. Bottom right figure is from <http://adg.stanford.edu/aa241/design/compaero.html>. Bottom middle figure is from <http://blog.adw.org/2010/06/is-the-church-a-cruise-ship-or-a-battleship/>. Top right figure is from <http://shamazkhan.wordpress.com/2011/01/10/agile-beams-active-electronically-scanned-array-radars/>. Bottom left figure is from <http://www.eece.ksu.edu/research/mars/transceiver.html>. Top middle figure is from <http://www.robotgear.com.au/Product.aspx/Details/376>
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- ▶ Page 9: Top right figure is from <http://www.takshak.org/iframes/circuitdebugging.php>, bottom right figure is from <http://melab.hanyang.ac.kr/common/read.asp?oper=2&num=7>.
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Questions ?